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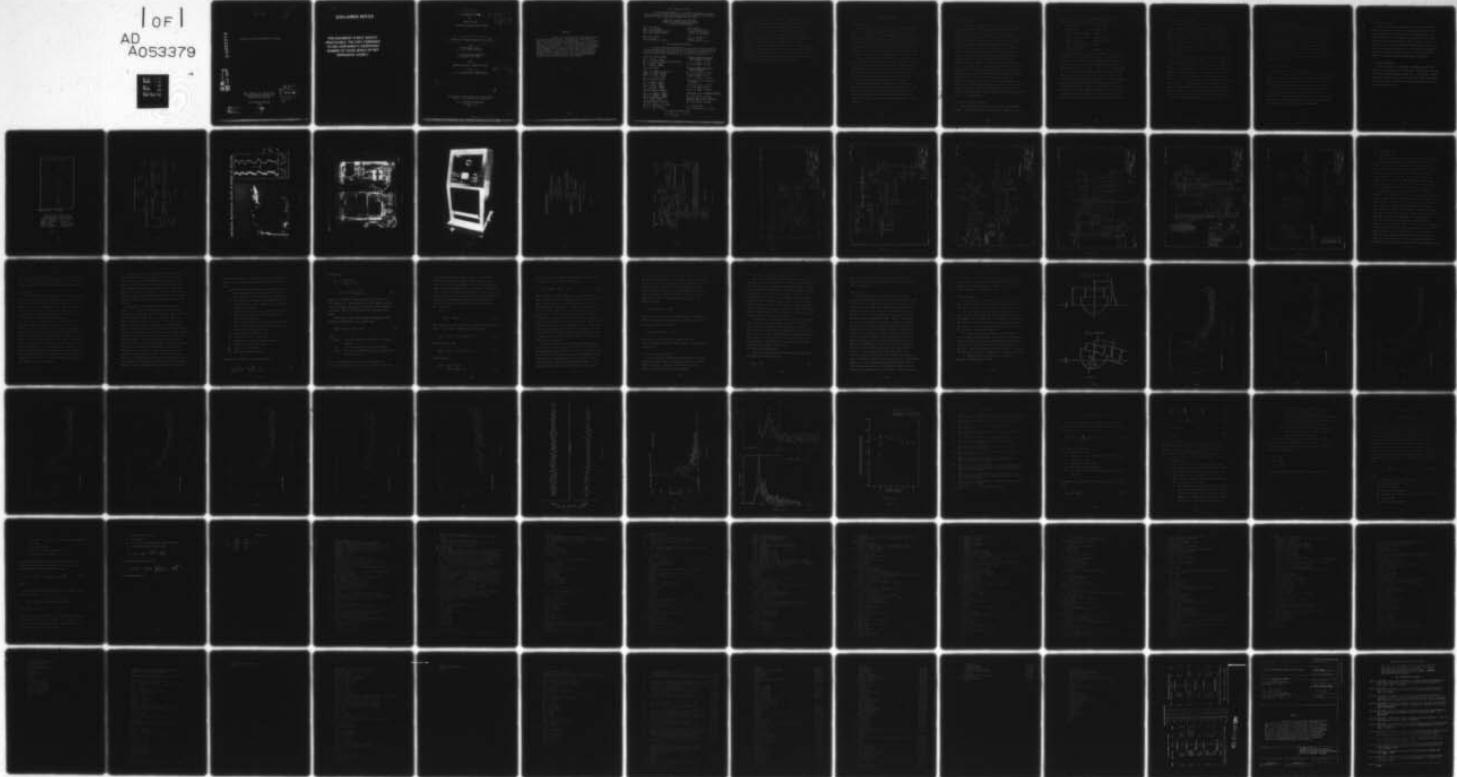
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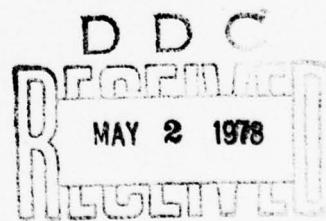
A REPORT ON SHIPBOARD WAVEHEIGHT RADAR SYSTEM

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9 FINAL TECHNICAL REPORT

Project SR-1234
"Shipboard Waveheight Radar System"

(12) 80p

A REPORT ON SHIPBOARD WAVEHEIGHT RADAR SYSTEM

by

(19) D./Chen and D./Hammond

U.S. Naval Research Laboratory
Washington, D.C. 20375

under

NAVSEA Work Request N00024-75-WR-51525

and

U.S. Coast Guard MIPR-Z-70099-4-43693

(14) SSC-SL-7-13

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ABSTRACT

A microwave shipboard wave-height radar sensor for measuring ocean wave spectra, developed by the Naval Research Laboratory, was installed on the containership *S.S. McLEAN*, February, 1975. The sensor's performance, design, and analysis of data for one data run are discussed. The radar system has a 3 centimeters wavelength, 2 nanoseconds pulse width, 100 watts of peak transmitted power, 10,000 pulse per second repetition rate, 2-foot parabola antenna diameter, 7 decibel receiver noise figure, 100 pulses per second equivalent pulse processing rate, and a 1-foot resolution. Results are in reasonable agreement with airborne measurements. Areas for improving the system are also discussed.

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Acknowledgments

The authors of this report wish to thank Kalen J. Craig for his indomitable spirit in developing the radar system and also the testing and evaluating it at sea. The assistance of James Kenney in preparing the equipment for use in the 1974-1975 season is greatly appreciated. Mr. Charles Buhler's skill in the assembly of the antenna package was an excellent example of assistance from the support services at NRL. Mr. E. A. Uliana's programming assistance made the data analysis possible. The cooperation from the personnel from the Teledyne-Ryan Corporation and the crew of the S.S. McLean made our task much easier. Without the help from any of these individuals, success in this effort may not have been possible.

SHIPBOARD RADAR

1.0 Introduction

The ability to measure the wave spectra in the open ocean from a moving vessel has met with varying degrees of success. Each sensor to date has suffered in its performance due to environmental conditions or due to its placement for measuring the unperturbed sea. This report will discuss the utilization of a microwave sensor on a moving vessel for measuring the open ocean wave spectra. Employing microwaves, some of the limitations of other sensors are not experienced.

Tucker [1] developed the Tuckermeter for measuring the wave spectra from a moving ship by sensing changes in water pressure due to surface wave conditions. The Tuckermeter is placed below the water line and thus requires calibration for each wave frequency, ship speed, and depth. Since the sensor operates on pressure, it performs as a low pass filter and will not sense the higher frequencies.

A microwave shipboard wave height radar sensor for measuring the ocean wave spectra was developed by the Naval Research Laboratory (NRL) and was installed on the S.S. McLean in February 1975 and its performance, design, and analysis of data for one data run will be discussed in this report.

2.0 Radar System

2.1 Introduction

Any sensor that profiles the ocean surface and measure the height variations can provide the necessary information for deriving the wave spectra. Since radar is a range measuring device, it lends itself ideally for this purpose. Profiling waves with a radar requires that the radar employ a very narrow antenna beam and very narrow pulses. The narrow antenna beam illuminates a small spot size on the ocean surface and the narrow transmitted pulse width permits resolving the fine height structure of the waves. Using a radar with these features on a tower, the radar returns clearly show the ability of the system to profile the ocean as shown in Figure 2-1. In addition the narrow antenna beam permits the radar to be aimed away from the nadir allowing measurements of the ocean unperturbed by the bow wake of the moving vessel. Microwaves also permit the radar to be operated day or night, in rain or fog, with bow spray or no bow spray, and continuously or intermittently. However, the radar is ineffective when a solid sheet of water splashes across the antenna beam. This situation occurs so infrequently that it can be discounted.

2.2 Radar System Parameters

The radar's high resolution of one foot is achieved by using a very narrow pulse of 2 nanoseconds. Figure 2-2 is a

functional block diagram of the radar system; and the principal characteristics of the radar system are:

Wavelength	3 centimeters
Pulse Width	2 nanoseconds
Peak Transmitted Power	100 watts
Pulse Repetition Rate	10,000 per second
Antenna Diameter	2-foot parabola
Receiver Noise Figure	7 db
Equivalent Pulse Processing Rate	100 per second

The R.F. components for the transmitter and receiver are mounted in a watertight enclosure on an antenna pedestal located about 90 feet above the ship's water line. The antenna is pointed abeam and tilted down and out about 15 degrees with respect to nadir. Figure 2-3 is a photograph of the antenna mounted on the starboard side of the ship's bridge and the figure to the right shows a sample of the measured data before processing. Figure 2-4 shows transmitter (upper box) and receiver (lower box) as they are mounted inside the watertight enclosure. The control and display circuits are located remotely from the transmitter-receiver assembly in a standard half rack as shown in Figure 2-5. All the timing and control signals are derived in this unit.

2.3 Principle of Operation

The 10 KHz timing generator, shown in Figure 2-2, triggers the transmitter and synchronizes the receiver signal processing. The R.F. transmissions consist of 2 ns wide pulses at 10 GHz carrier frequency with a peak power of 100 watts and with a pulse repetition rate of 10,000 per second. The reflected R.F. signals from the ocean is amplified in the receiver to a usable level. Employing an envelope detector on the amplified signal results in a 2 ns wide video pulse.

Processing the 2 ns wide pulses requires circuitry in the system with bandwidths of 500 mHz. It is desirable to operate at a lower bandwidth where components are more easily used and obtained. By employing a sampling scope for display and signal processing, it is possible to make this bandwidth transformation which is equivalent to a video pulse that is 200 microseconds wide or a bandwidth of 5 KHz. Thus the use of standard low speed logic circuits can be used for signal processing resulting in a simpler and more reliable system. The principle of operation of a sampling scope is well known and will not be discussed. See references [2, 3, 4] for the particular scope used in this radar.

2.3.1 Automatic Gain Control

The block diagram for the automatic gain control (AGC) is shown in Figure 2-6. The amplitudes of the returned pulse is changed due to (1) scattering from a rough surface, and/or (2) large changes in the viewing angle. The effect of the scattering is to induce rapid changes in pulse amplitude while the changes due to viewing angle are much slower and are on the order of tens of seconds. The time constants in the AGC loop are adjusted to compensate for the slow changes of ship's roll but will not affect the rapid pulse-to-pulse changes.

2.3.2 Range Tracker

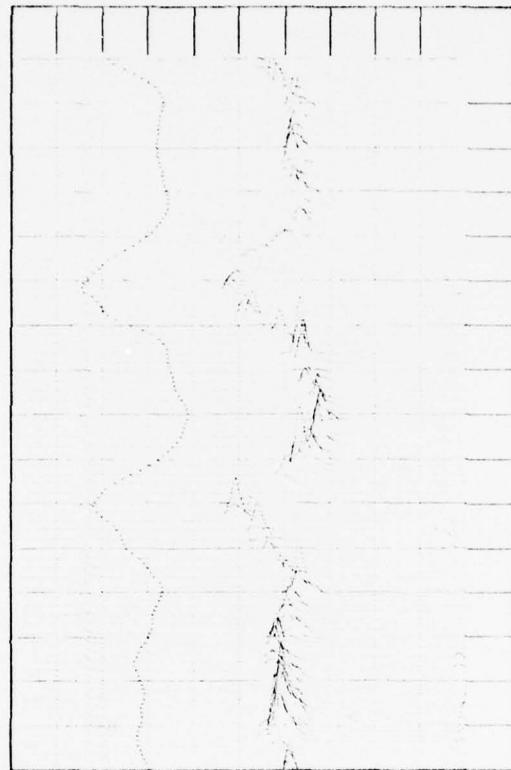
The block diagram for the range tracker is shown in Figure 2-7. The remote programming feature of the Tektronix 3T5 sweep unit [4] is used in the range tracking loop. Employing this feature permits the range tracker to automatically adjust the time delay of the scope trigger so that the returned video pulse always remains centered on the sampling scope screen. The details of the tracker are shown in Figures 2-8, 2-9, 2-10, and 2-11. The signal flow and logic controls are quite involved and will not be discussed here. This information will be provided if requested.

2.3.3 Output Signal

The range tracker follows the peaks and crests of the waves and produces an output voltage proportional to the range excursions. The voltage to be recorded changes one volt for each 12.8 feet change in the distance of the radar antenna from the ocean surface. When as many as fifty successive range pulses have been missed, indicating loss of range lock, an error light is lit. The AGC meters shows the value of attenuation inserted in the receiver amplifier which is a measure of the returned signal amplitude.

2.4 System Drawings

The drawings of the system are included so that the details of the system can be trace out. The drawings include a simplified block diagram, Figure 2-8, a total radar system schematic, Figures 2-9, 2-10, and 2-11, the cable connections and control panel, Figure 2-12, and the transmitter-receiver schematic, Figure 2-13.



WAVE STAFF RADAR

PULSE DATA FROM THE
CHESAPEAKE LIGHT TOWER
WAVE HEIGHTS 5 FEET
HOR. SCALE 2.5 FT/DIV
VERT. SCALE 1 SEC/DIV

Figure 2.1

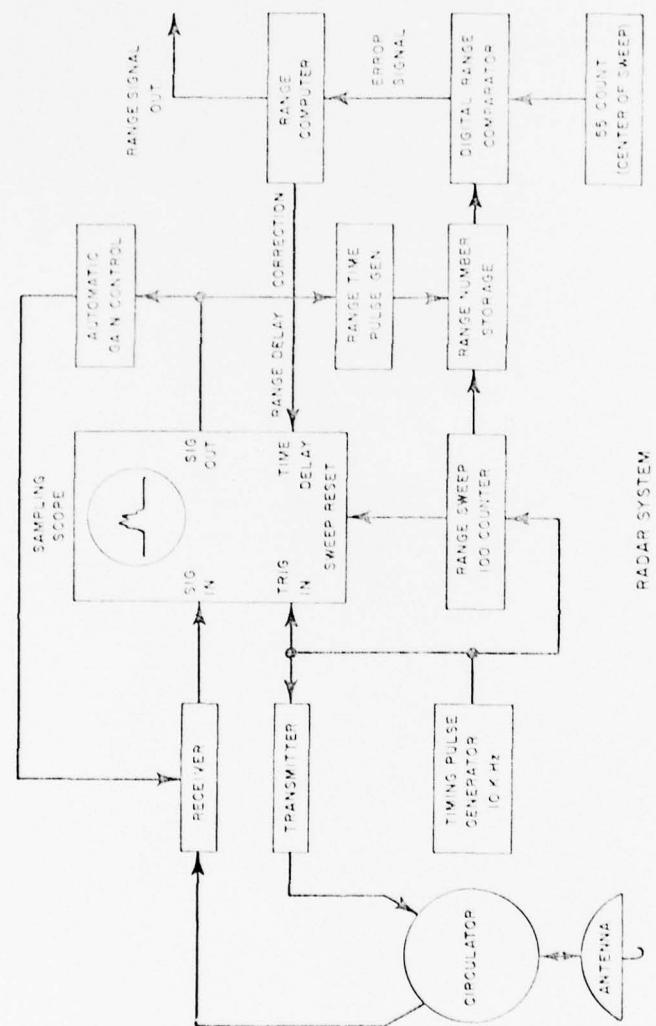


Figure 2.2

SHIPBOARD WAVEHEIGHT RADAR ON-BOARD SS - McCLEAN

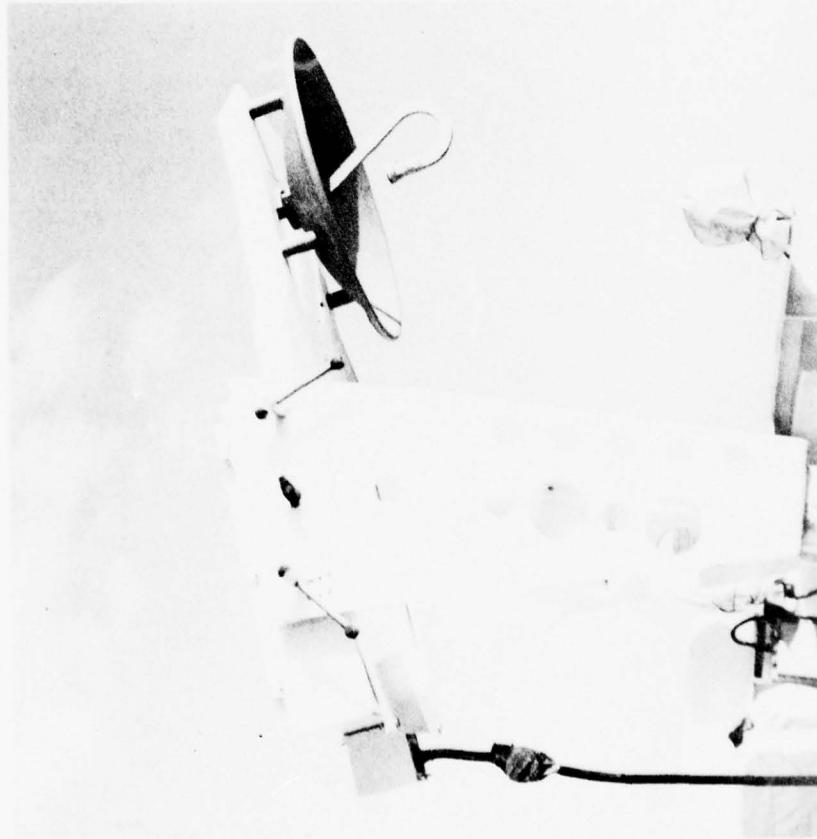
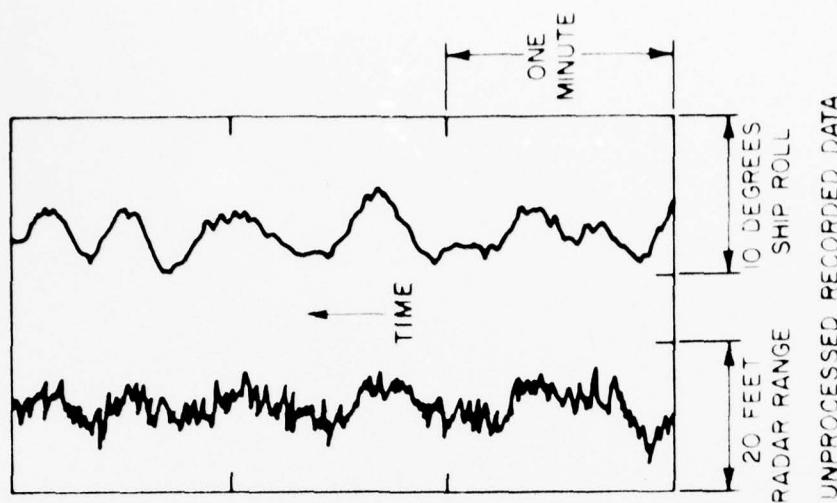


FIGURE 2-3

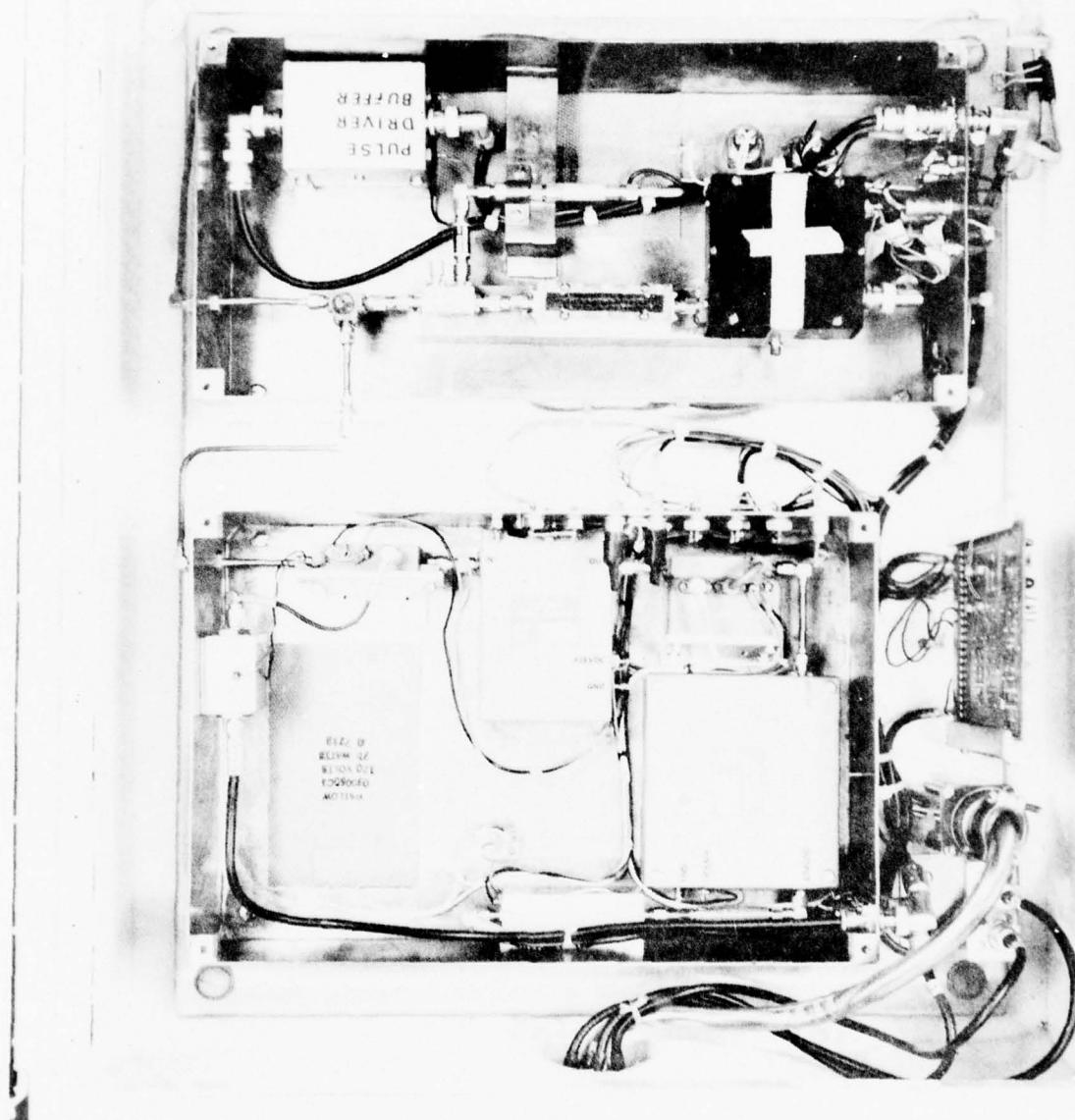


FIGURE 2-4

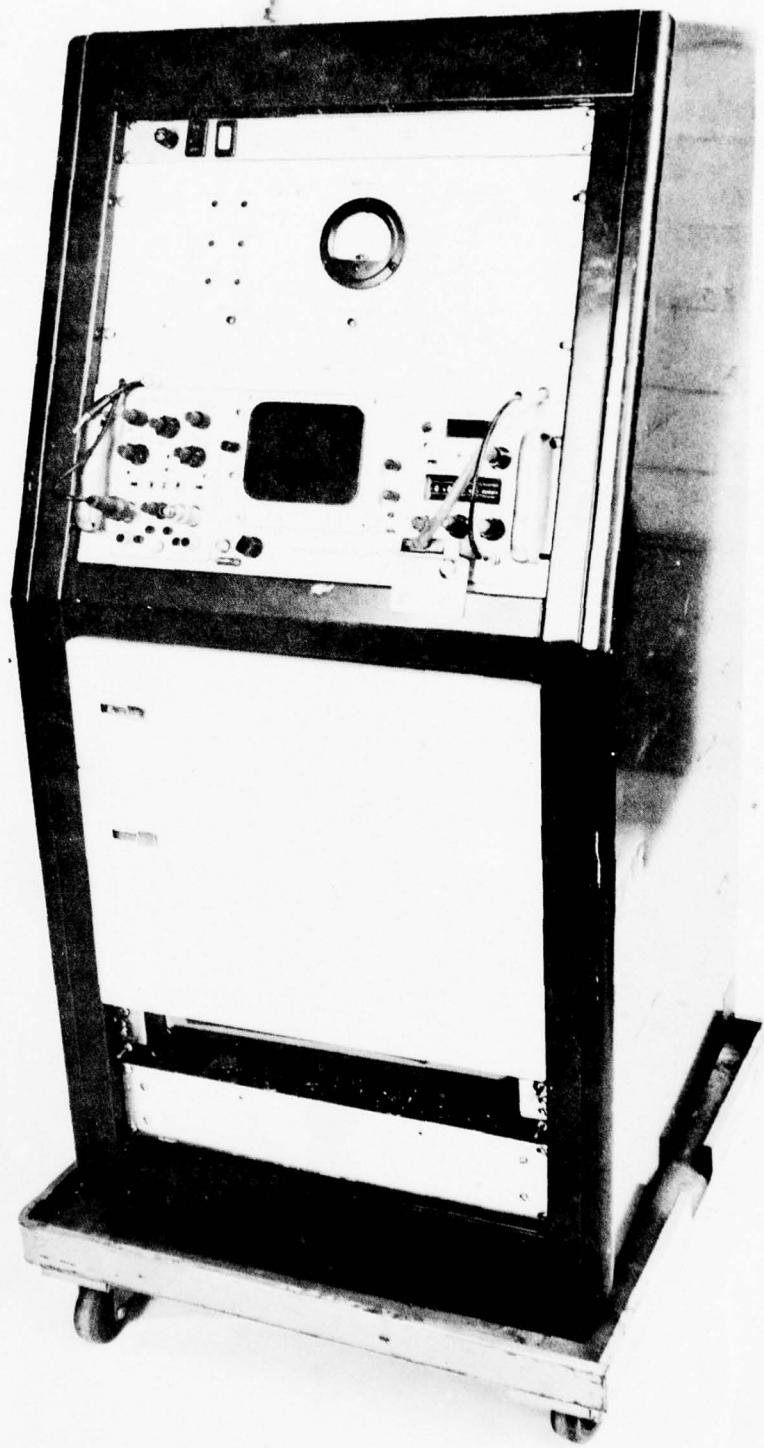


Figure 2-5

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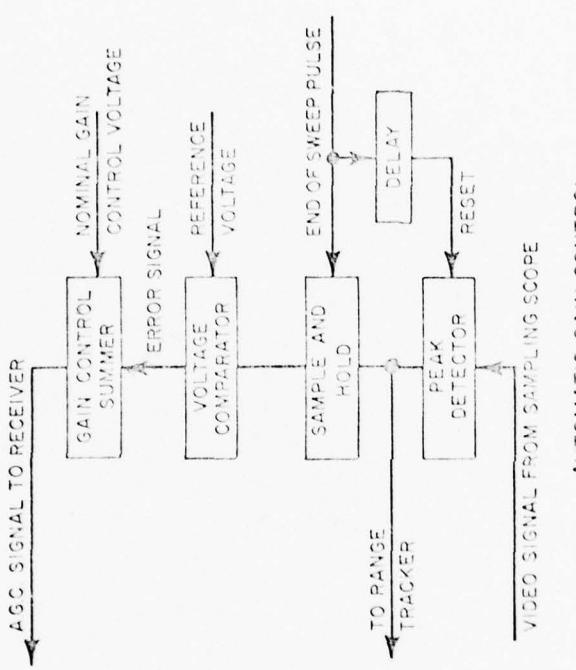


Figure 2.6

AUTOMATIC GAIN CONTROL

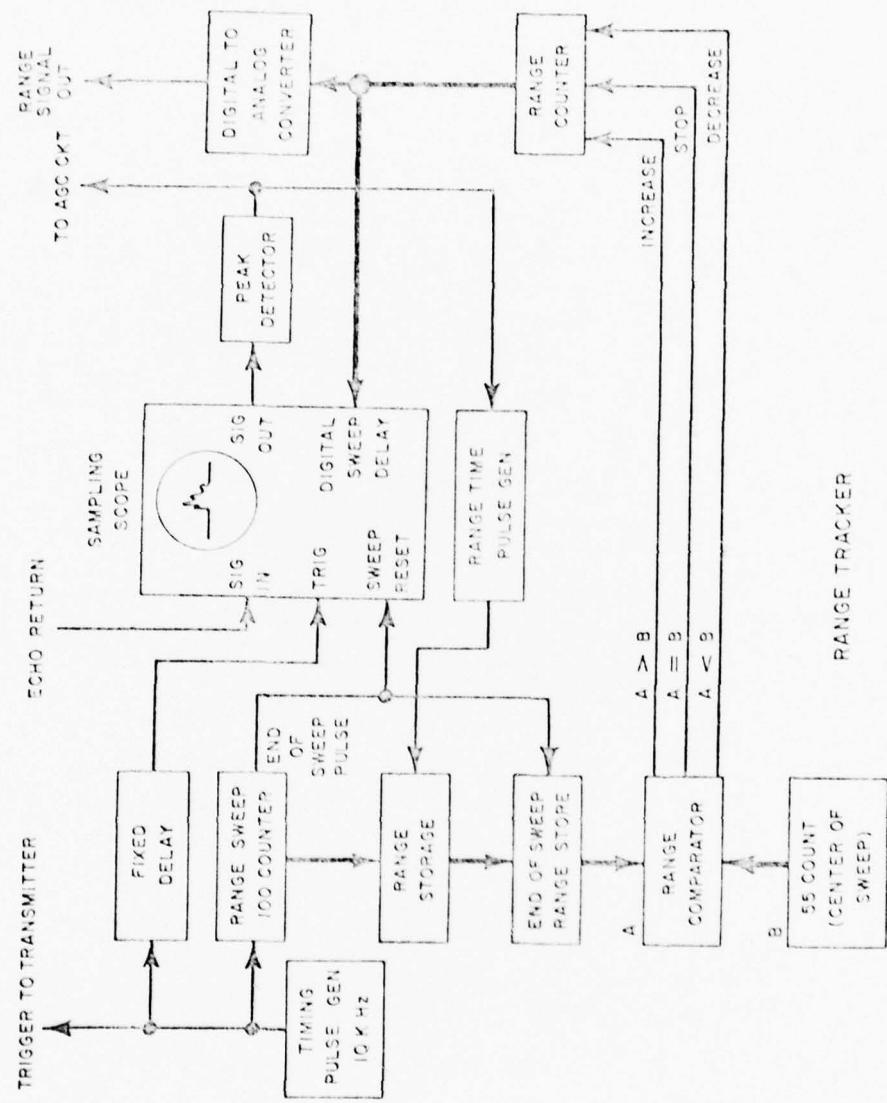


FIGURE 2.7

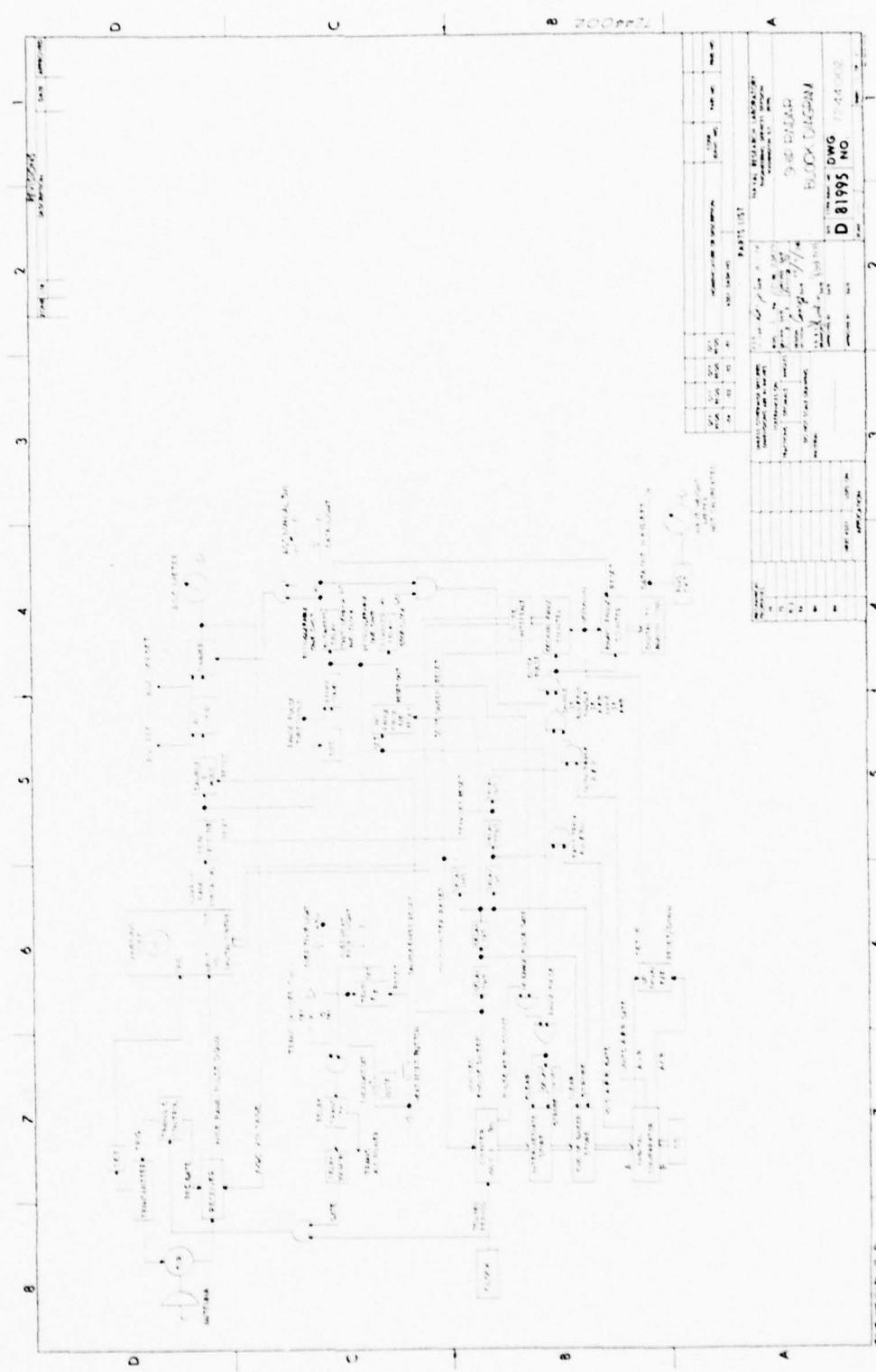


Figure 2.8

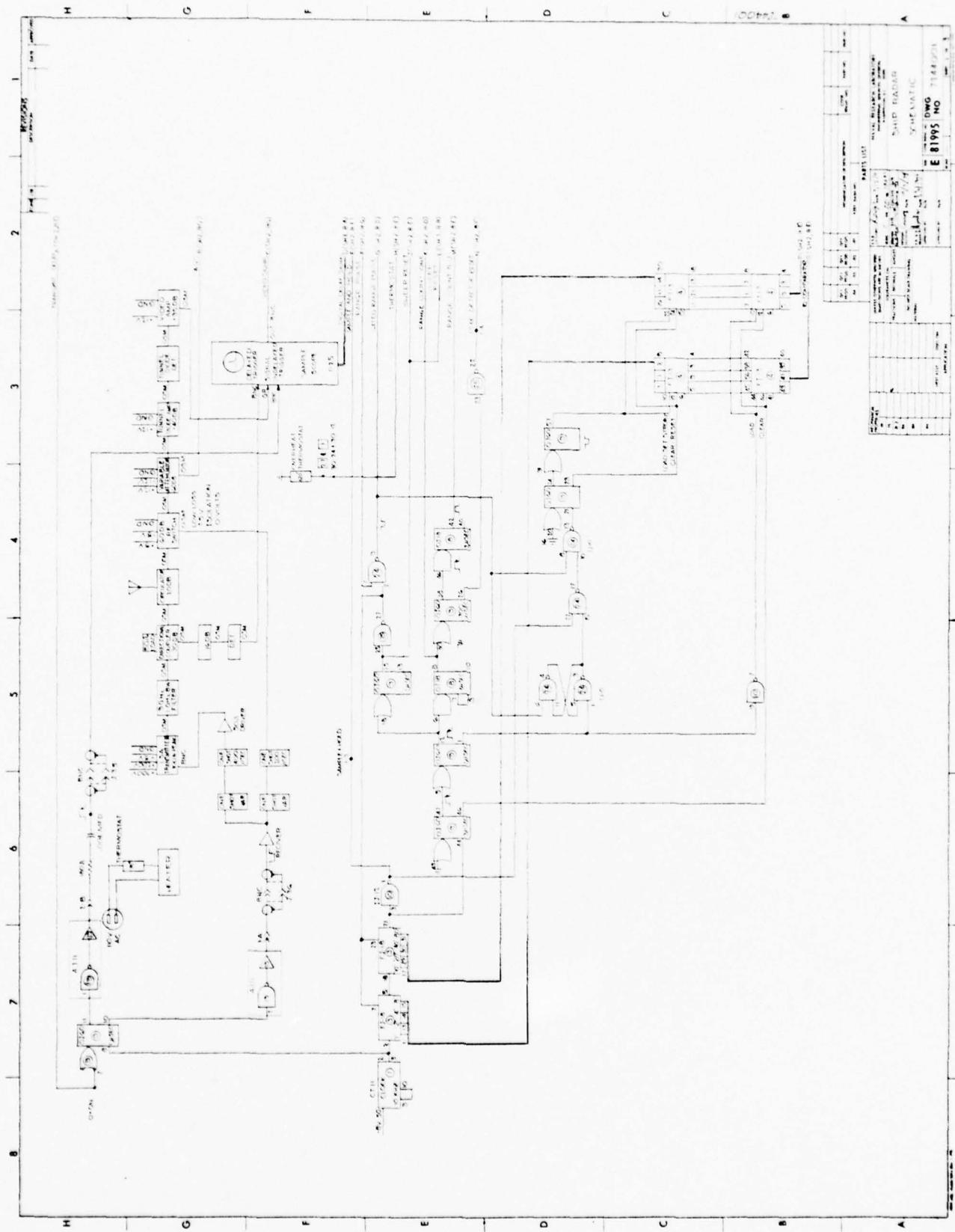
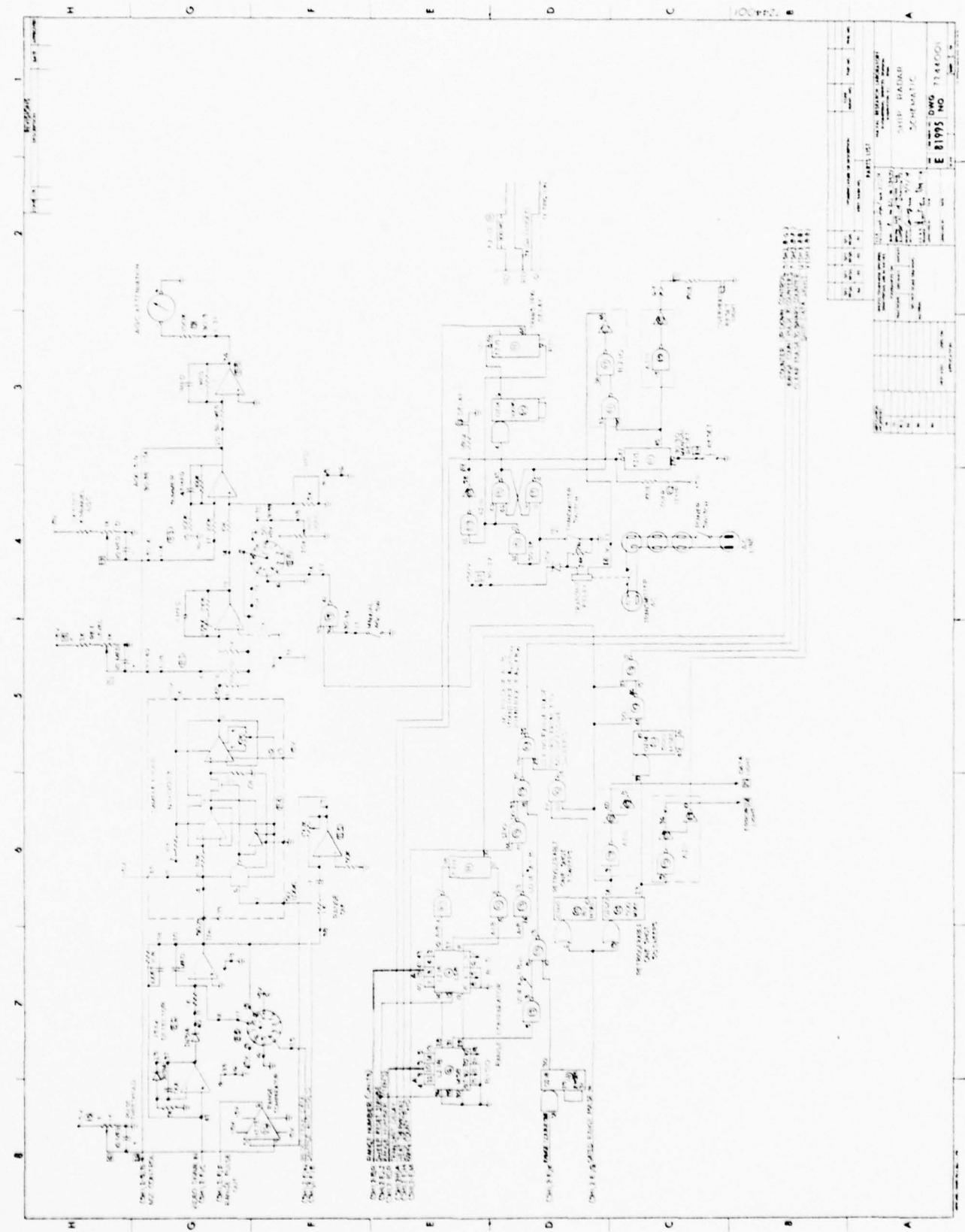


Figure 2.9



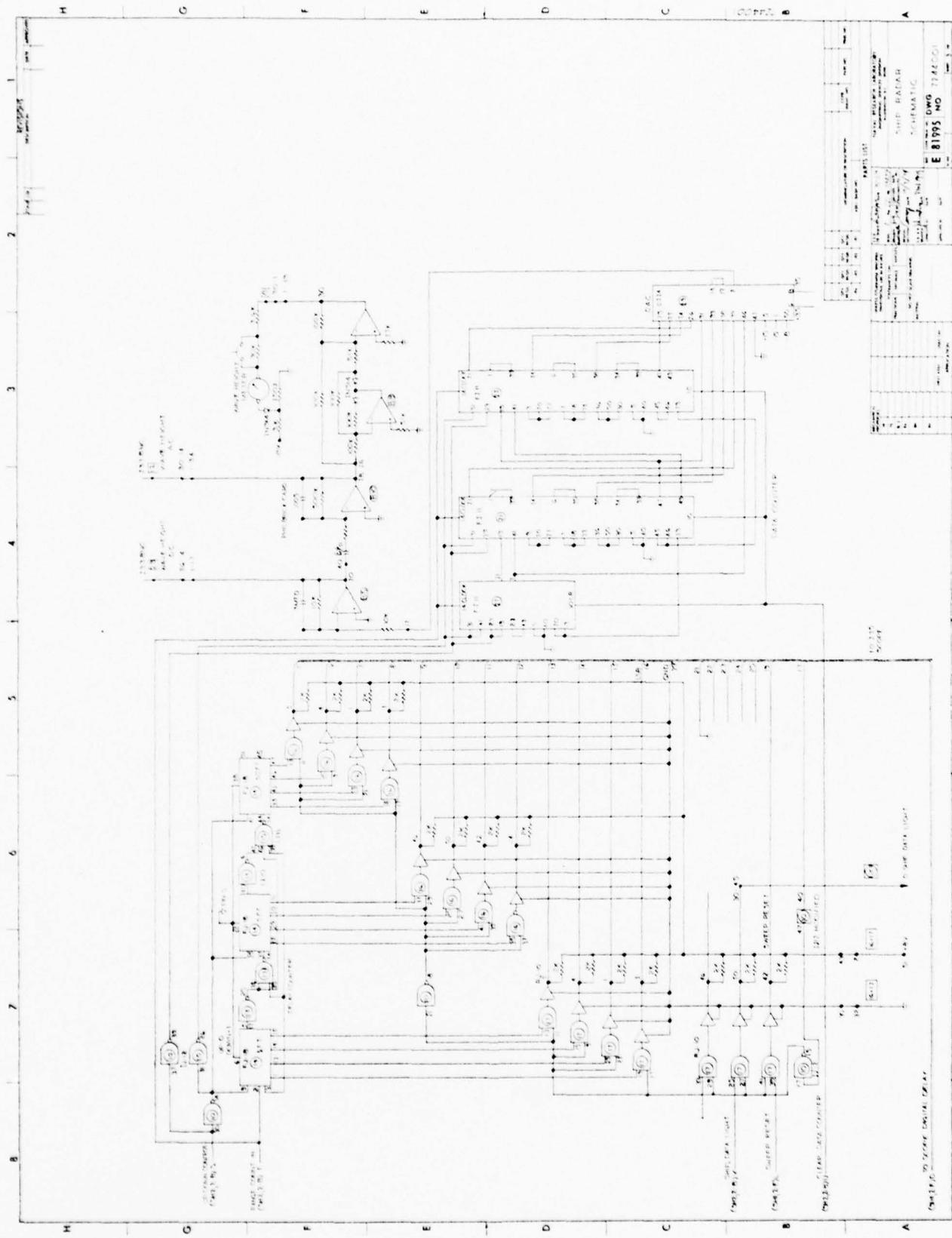


Figure 2.11

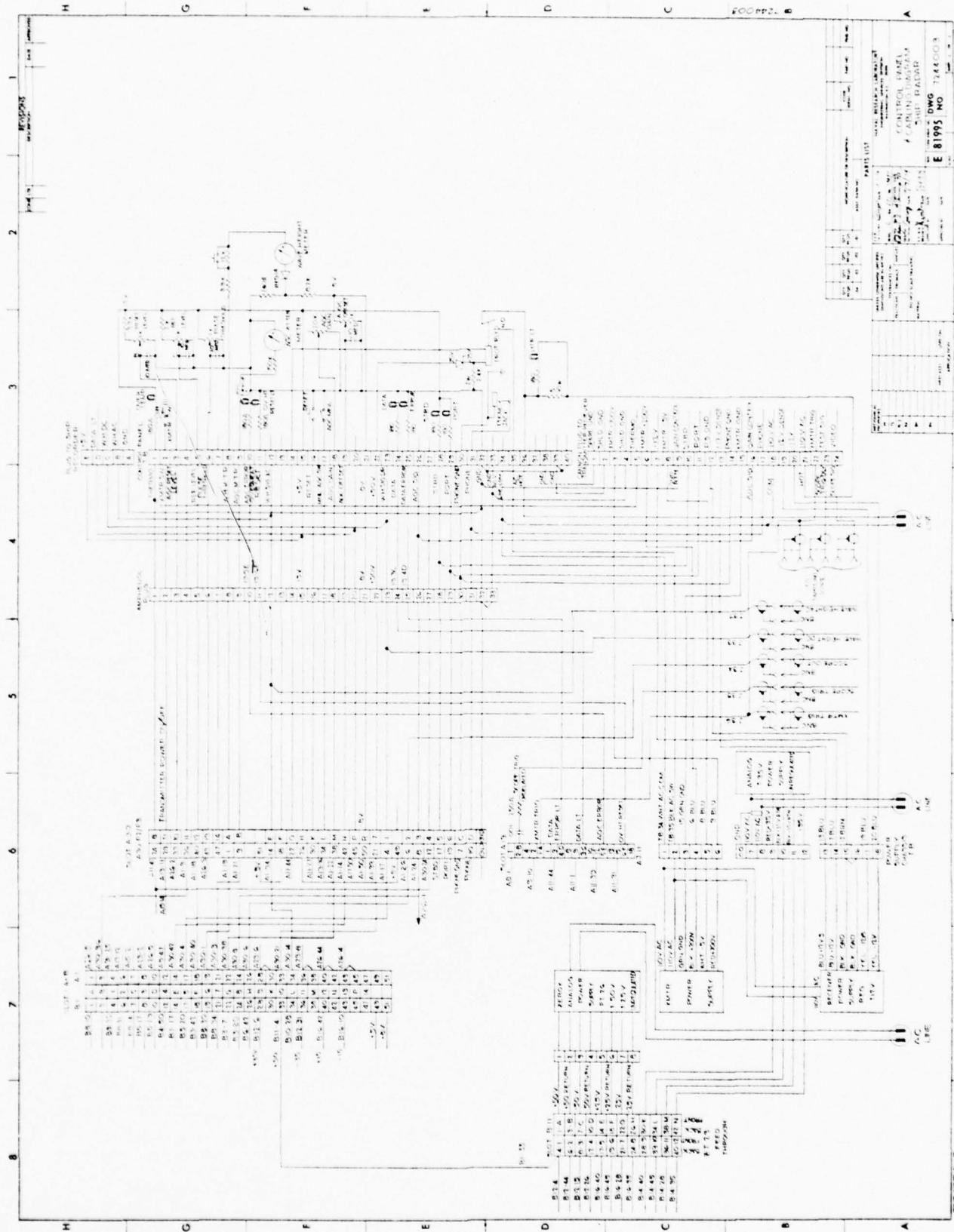


Figure 2.12

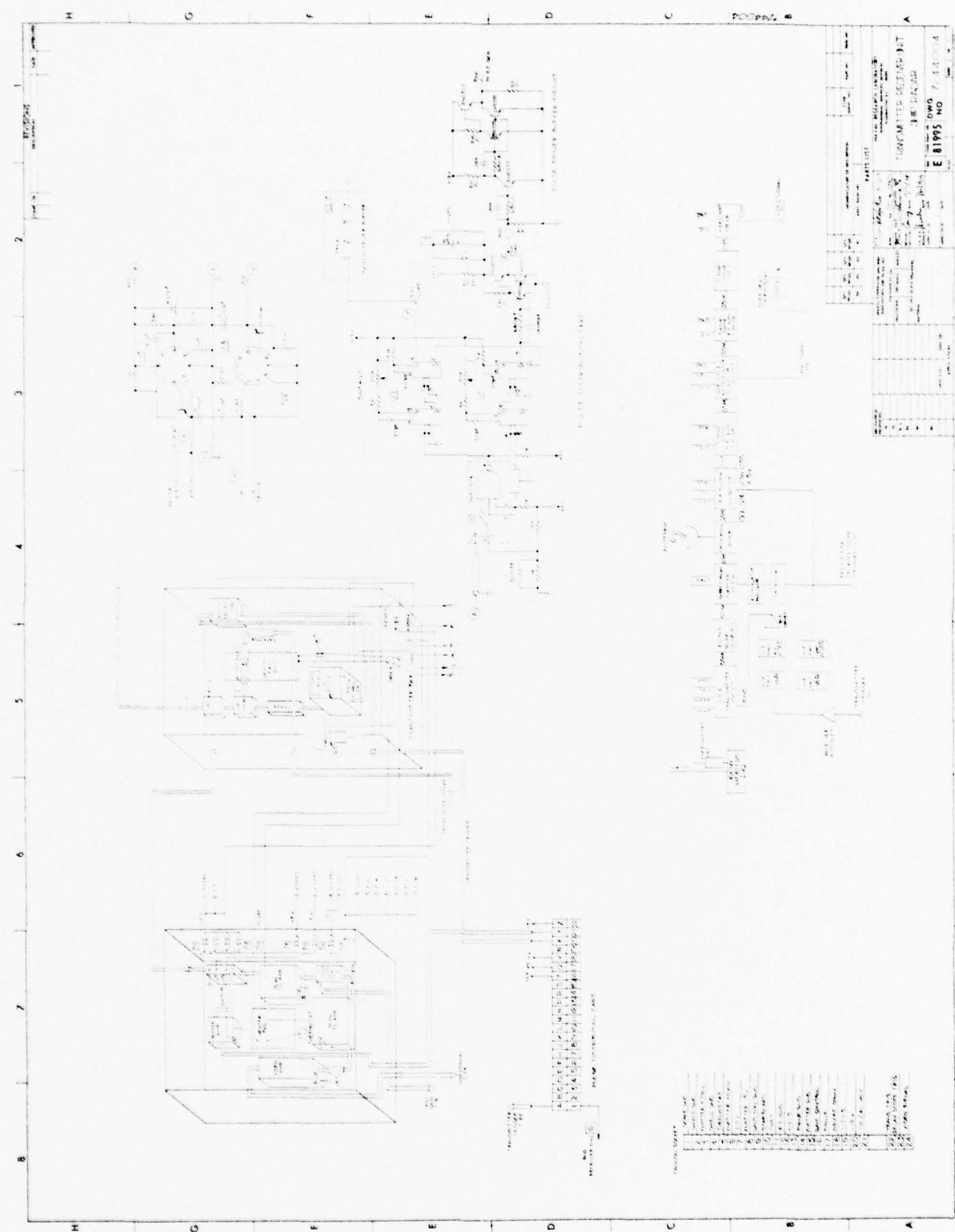


Figure 2.13

3.0 Data Reductions

3.1 Experimental Test

The nanosecond radar is located starboard on the bridge of S.S. McLean and adjusted to view the ocean at a look angle of 15° from nadir and away from the bow wake. On February 6, 1975, while the ship was underway from Elizabeth, New Jersey to Portsmouth, Virginia, simultaneous ocean surface data was taken by the shipboard radar while an airborne laser profilometer and airborne nanosecond radar were measuring the same seas. The object of this effort is to establish the validity of the shipboard measured data.

The shipboard radar data was recorded at 30 minute intervals starting at 8:20 A.M. EST on February 6, 1975 and ended at approximately 2:00 P.M. of the same day. Each file of data started with about a minute of zero level setting and followed by a minute of calibrations. The ship was travelling at 29 knots at 214° heading. Approximately 9:26 A.M. the aircraft intercepted the ship's track at 38°3' N in latitude and 74°41' W in longitude. Airborne data was recorded at 500 feet and at 1000 feet altitude in the immediate vicinity of the ship's path. At 500 feet altitude, a laser profilometer [5] and an airborne nanosecond radar were used to profile the ocean surface. At 1000 feet altitude, NRL's airborne nanosecond radar was operated in the wave spectrometer mode. Aircraft data was recorded continuously at intervals of about 90 seconds as the aircraft

was flying at 146 knots ground speed. Exact coincidence of data taking was not possible and comparison of the data is based on approximate times from the ship and aircraft logs.

3.2 Analysis

The dynamics of the two platforms (aircraft and ship) while recording ocean surface data was significantly different. In addition to their peculiar motion characteristics, the data was recorded in different manners. One, continuously in analog and the other, intermittently and digitally. To effect a better comparison of the data, it was necessary that the data be reduced to some common base.

The analog shipboard data was recorded in real time, and then sampled and digitized at 8 Hz off-line. This digitizing rate was a compromise, taking into account the longest ocean wavelengths expected to be experienced by the ship and the storage capacity of in-house computers to process the data. The aircraft data was digitized at 90 Hz rate in real time but was utilized at 15 Hz rate off-line. Again a compromise was made with the high frequencies which were not overly significant being truncated and at the same time trying to maintain equal spatial resolution with the shipboard data. The data storage and handling capacity of the computer was also of concern.

The largest amount of dynamic motion or movement other than forward velocity of a ship is its roll. In Figure 2-3 was shown a sample of the shipboard radar output and the output of the roll sensor. The magnitude and the period of the roll which affect the determination of the true ocean wave spectra must be removed before any real analysis can be conducted. However other motions also need to be corrected for unless they become second order in magnitude and can be ignored.

The geometry of the radar aboard ship is shown in Figure 3-1. The upper figure shows the direction of the ship moving into the paper and in still water thus remaining in the upright position. The radar measures $R(t)$, the distance from the radar to the surface of the water at an angle α with respect to the vertical. As the ship moves, the radar distance, $R(t)$, changes with the peak and crest of the waves. It is the magnitude and frequency of these radar distance variations that yields information for determining the ocean spectra. As mentioned earlier, ship motions, particularly roll, affects the magnitude of these variations yielding an erroneous wave height change. In the lower illustration of Figure 3-1 is shown an instantaneous roll position that the ship can assume. In this situation, the change in radar distance $R(t)$ is not due to waves but is due to the ship's motion and the radar then measures the distance R_0 instead of R_o . The magnitude of the change in R_o to R_0

needs to be determined. The geometry, assuming rigid body motion, for determining this change is shown in Figure 3.1 where

H_0 = distance from radar antenna to water surface, in the direction of ship's symmetric axis, and 0° roll angle; 76 ft. for H_0 in these calculations.

H_θ = distance from radar antenna to water surface in the direction of ship's symmetric axis, caused by θ degrees roll angle.

L_0 = horizontal distance of radar antenna from center of gravity of the ship at 0° roll angle, 40 ft. for these calculations.

θ = angle of roll; positive in clockwise convention, from vertical upright position, by looking into the direction of ship's heading.

R_0 = radar distance of 0° roll angle,

R_θ = radar distance at θ° roll angle,

α = look angle of radar antenna, 15 degrees,

CG = center of gravity of the ship,

MC = metacenter of the ship,

BC = buoyant center of the ship.

From the Law of Sine's, it can be shown that

$$\frac{R_\theta}{\sin(\frac{\pi}{2} - \theta)} = \frac{H_\theta}{\sin(\frac{\pi}{2} + \theta - \alpha)} \quad . \quad (1)$$

Rearranging

$$\begin{aligned}
 R_\theta &= H_\theta \frac{\sin(\frac{\pi}{2} - \theta)}{\sin(\frac{\pi}{2} + \theta - \alpha)} \\
 &= (H_o - L_o \tan\theta) \frac{\cos\theta}{\cos(\theta - \alpha)} \quad (2)
 \end{aligned}$$

where R_θ is the radar distance to the water surface at a roll angle of θ° . Using Equation (2) the values for $R_\theta(t)$ can be determined from H_o , L_o , α , which are given and the roll angle θ which can be obtained from the ship's roll sensor.

Using $R_\theta(t)$ in the radar distance measurements, the following relationship can be established:

$$R_R(t) = R_\theta(t) + \zeta(t) + \Delta R(t) \quad (3)$$

where

$R_R(t)$ - the radar distance measurement to the water surface

$\zeta(t)$ - the instantaneous apparent wave height

$\Delta R(t)$ - the radar distance changes due to ship motion other than roll.

This relationship shows the effect of ship motion in conjunction with wave motion and their effect on the radar

range measurements, but assumes that any flexure which changes the distance between the radar antenna and the center of gravity is negligible. All other motion changes such as yaw, pitch and heave, etc. are combined into $\Delta R(t)$. $\zeta(t)$, the term of interest in describing the sea surface, is small in magnitude and it modulates the distance $R_R(t)$ which is large. The shipboard equipment was designed only to record this modulation and as a result a large distance bias, D , remains.

Let

$$R_A(t) = R_R(t) - D \quad (4)$$

where $R_A(t)$ is called the relative radar range measurement and D is a constant. Substituting (3) into (4),

$$R_A(t) = R_0(t) + \zeta(t) + \Delta R(t) - D. \quad (5)$$

Rearranging the terms

$$R_A(t) - R_0(t) = \zeta(t) + \Delta(t) - D. \quad (6)$$

Redefining (6)

$$\begin{aligned} R_1(t) &= R_A(t) - R_0(t) \\ &= \zeta(t) + \Delta R(t) - D, \end{aligned} \quad (7)$$

where $R_1(t)$ is the relative radar range without the effect of ship's roll. Rearranging the terms

$$\zeta(t) = R_1(t) - [\Delta R(t) + D] . \quad (8)$$

This results in $\zeta(t)$ describing the sea surface variations. There still exist $\Delta R(t) + D$ which has not been accounted for because D is basically a DC term. This however can be removed by filtering the data. $\Delta R(t)$, as defined, consists of all the other ship motions affecting the radar range measurements. The high frequency components of $\Delta R(t)$ are of such low magnitude that their effect on the wave measurements is negligible, whereas the low frequency component can be effectively removed with a high pass filter. $\zeta(t)$ still contains a doppler term and it must be taken care of before it is possible to study the ocean surface characteristics. In Appendix A the filtering process employed to remove the $\Delta R(t)$ term is discussed.

The term $\zeta(t)$ describing the amplitude variations of the waves as the radar profiles the surface while the ship is underway includes a doppler term. Due to the velocity of the ship and the doppler effect, the waves encountered are foreshortened. Thus the wavelengths measured are not the true ocean wavelength but an apparent wavelength. If the radar's instrumentation had incorporated a coherent R.F.

signal, the effect of the ship's velocity can, by appropriate instrumentation, be subtracted directly. Since this is not the case, it is necessary to correct the apparent spectra to a true spectra. In order to accomplish this, it is no longer possible to operate in the time domain but one must resort to working in the wave number or frequency domain. Thus the time function must be transformed into a wave number function:

$$\zeta(t) \text{ transforms } \rightarrow \psi_A(k_A)$$

where k_A is the apparent wave number because of the doppler effect and $\psi_A(k_A)$ is the apparent wave number spectrum. It can be shown, Appendix B, that

$$\psi_A(k_A) \text{ transforms } \rightarrow \phi(\omega_T)$$

where $\phi(\omega_T)$ is the true wave frequency spectrum. $\phi(\omega_T)$ is the spectrum shown in the results. Equation (B-2) in Appendix B.

3.3 Fast Fourier Transform

A fast fourier transform [6] was used in the spectral analysis and since this is standard operation, it will not be discussed here. However it was necessary to use a Hamming function [7] to smooth the spectra.

3.4 Spectral Bandwidth Vs. Spectral Resolution

The 2-foot parabolic antenna illuminates a footprint [8] about 3.11 feet in diameter at a radar range of 80 feet. This footprint size can only resolve wavelengths larger than six feet which is equivalent to an apparent cutoff frequency of 0.908 Hz. The true cutoff frequency is 0.358 Hz after removing the Doppler effect. Translating this true cutoff frequency into a true wave period results in wave period of not less than 2.8 seconds. The high pass filter, discussed earlier, removes wave periods in the data larger than 7 seconds. The wave spectra shown in the results are calculated for the wave period window of 2.8 to 7 seconds.

If different spectral bandwidths are desired, the upper frequency bound can be raised by increasing the antenna size. The lower frequency bound can also be extended but this requires removing the ship's motions without the use of high pass filter. The analysis requires ship motion sensors at the site of the antenna to record the actual excursions of the antenna. In this manner it is possible to resolve the longer wavelengths of the spectra.

3.5 Significant Wave Height

Significant wave heights were determined from Equation (9) employing the following [9]

$$H_{1/3} = 4\sqrt{E} \quad (9)$$

where $H_{1/3}$ is the significant height and E is the total energy of the waves, which is obtained by integrating the wave spectrum curve.

3.6 Results

Wave spectra from the shipboard measurements are presented in Figure 3-2 to 3-9. The significant wave heights for files 1 to 8 of the data, respectively, are 6.90, 7.29, 7.33, 7.11, 7.10, 6.53, 6.53, and 6.37 feet. Since the sea was not in steady state conditions and fetch-limited seas did not exist, it is not possible to make comparison with the Pierson-Moskovity spectrum [10]. Airborne measurements are presented in Figure 3-10 and 3-11. Figure 3-10 shows the measurements made by laser profilometer and the nanosecond radar. Figure 3-11 shows the analyzed results from the airborne measurements. Laser profilometer registered 6.04 feet and the nanosecond radar registered 4.70 feet as the significant wave height based on data while flying at 500 feet. At 1000 feet, operating the nanosecond radar in wave spectrometer mode yielded significant wave height of 6.60 feet. All airborne measurements were made in the vicinity of the ship during part of the time the shipboard radar was recording data that is shown in file 2. Accordingly, Figure 3-11, the airborne measurements and Figure 3-3 of the shipboard measurements are combined in Figure 3-12 for comparison. The shapes are very similar. The significant

wave heights of all the measurements shown in Figure 3-13. The data are in reasonable agreement, especially when time coincidence of the data is not possible and the aircraft covers such large area about the ship.

3.7 Discussions

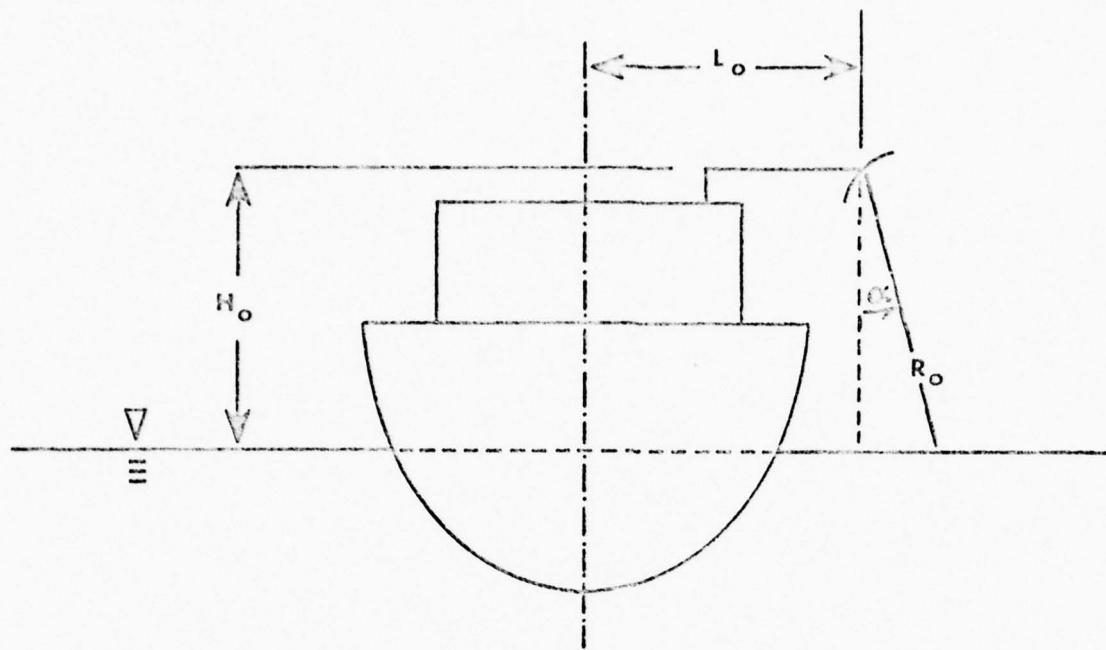
In any system, one can always find areas for improvement, and this radar is no different. After conducting the analysis of the data, several points should be noted.

1. Future radars for this purpose should record the total range as well as the range modulation by the waves. The advantage is to enable one to make absolute corrections for the ship motions; otherwise, only relative corrections can approximately be made for ship motions, and still leave the DC offset as an unknown quantity.

2. A shipboard radar measurement of wave spectra permit viewing the undisturbed area of the sea. With better time and spatial resolution and by employing accelerometers on the antenna, the ship motion effects can be removed directly. Three accelerometers and three angular sensors at the site of the radar are recommended for future measurements.

3. Shipboard radars can operate in all types of weather, twenty-four hours a day.

(A) VERTICAL UPRIGHT POSITION



(B) ROLL POSITION

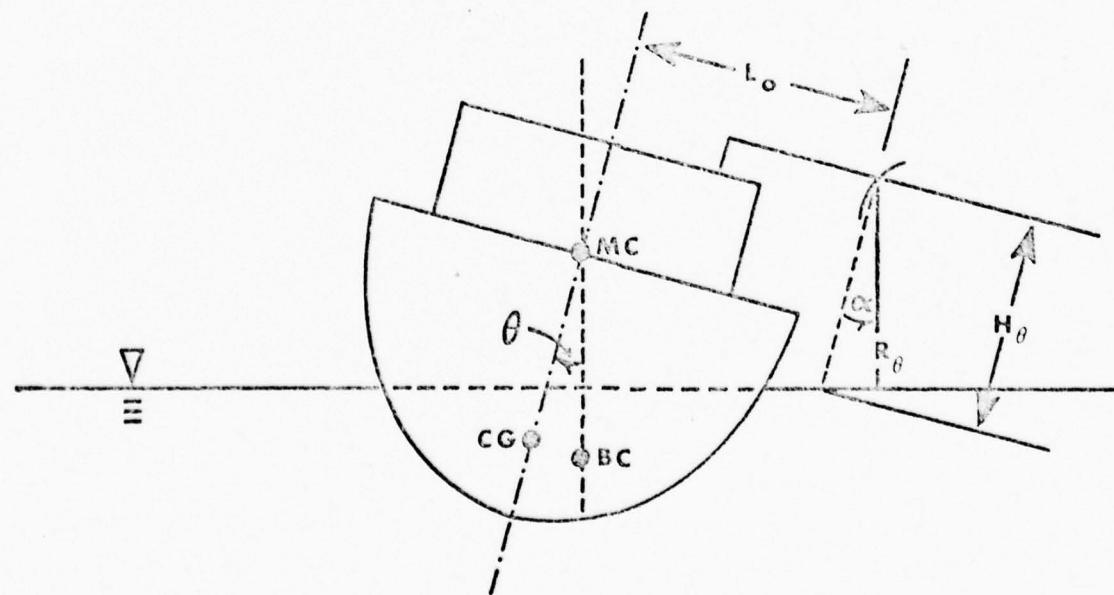


Figure 3-1

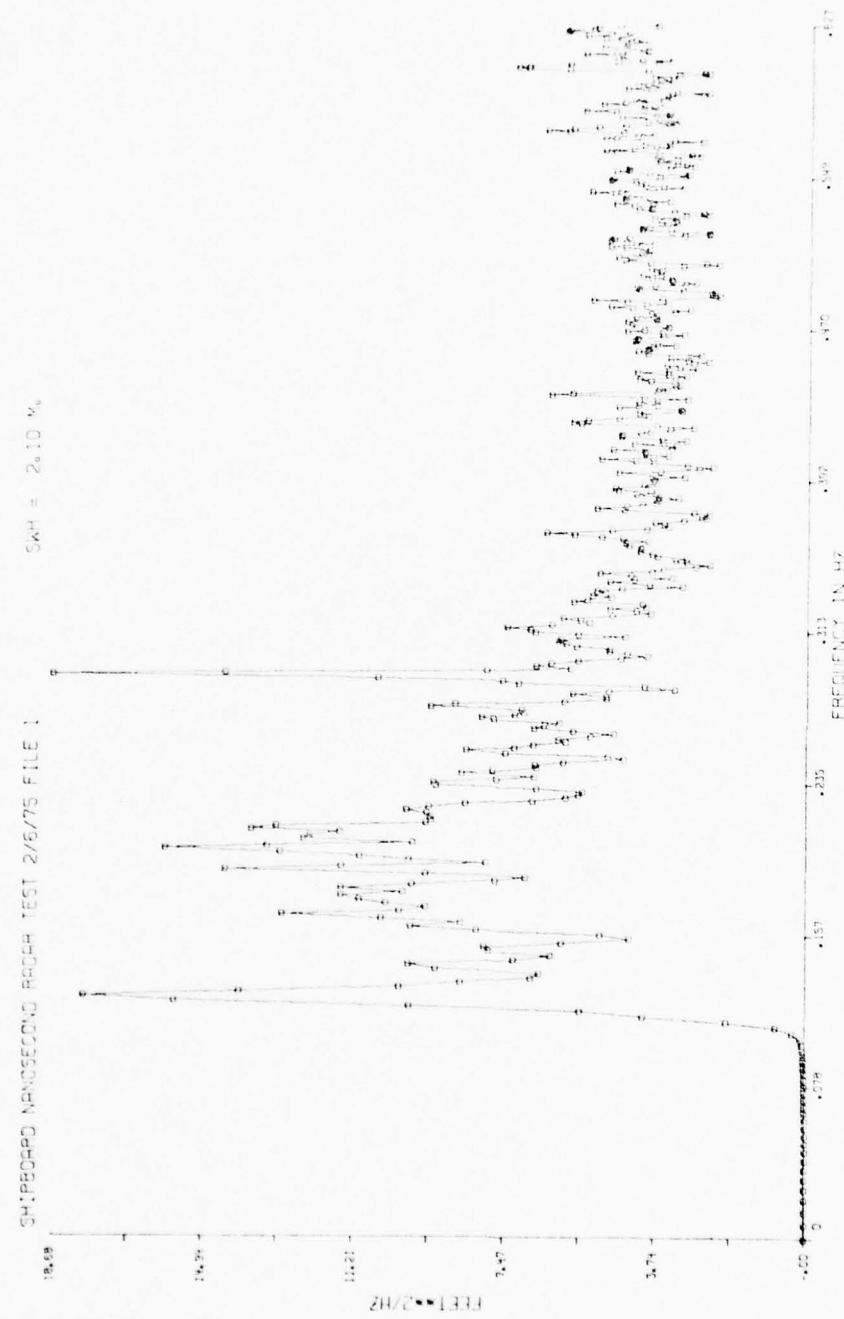


Figure 3-2

Figure 3-3

Figure 3-4



Figure 3-6



Figure 3-8

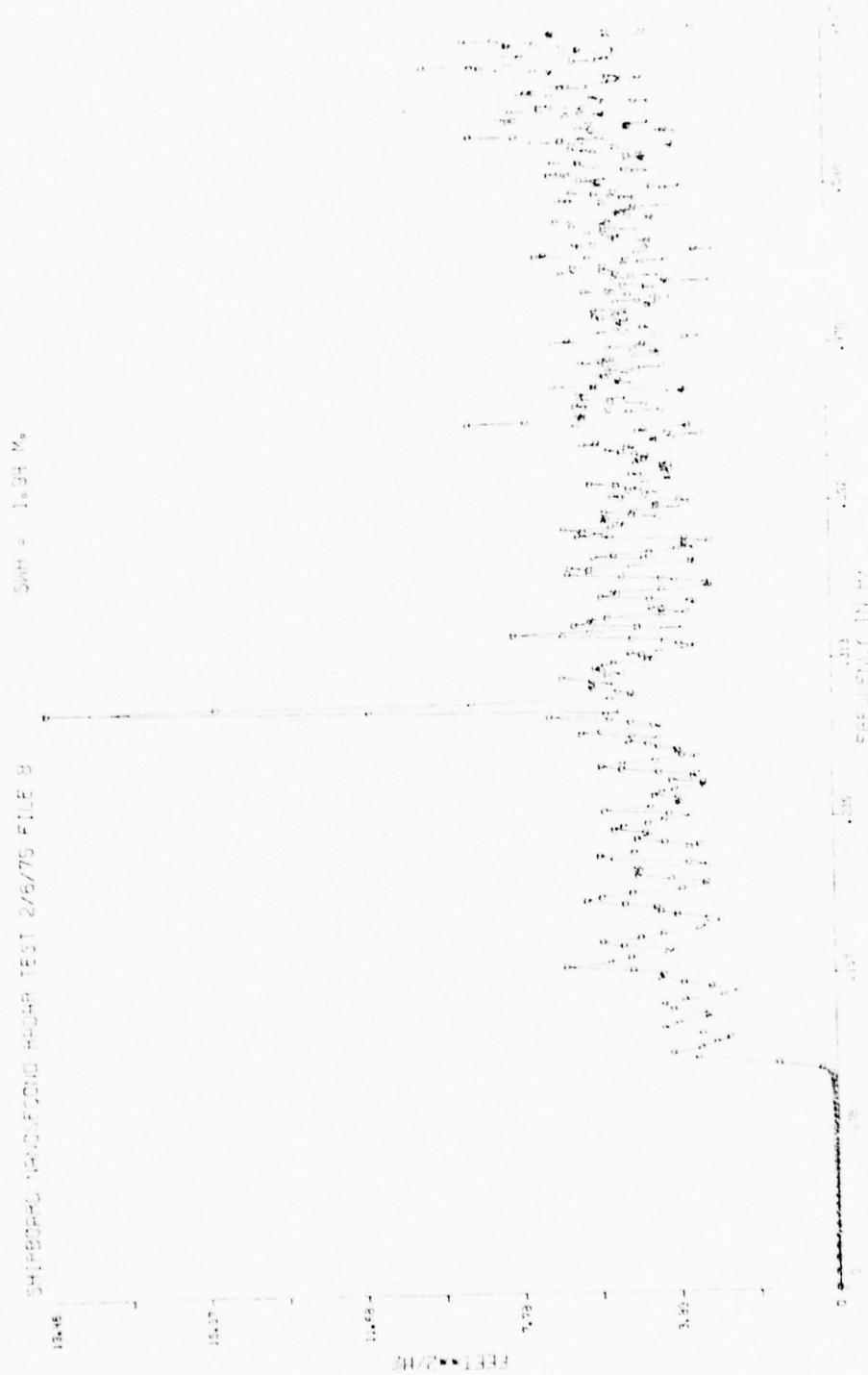


Figure 3-9

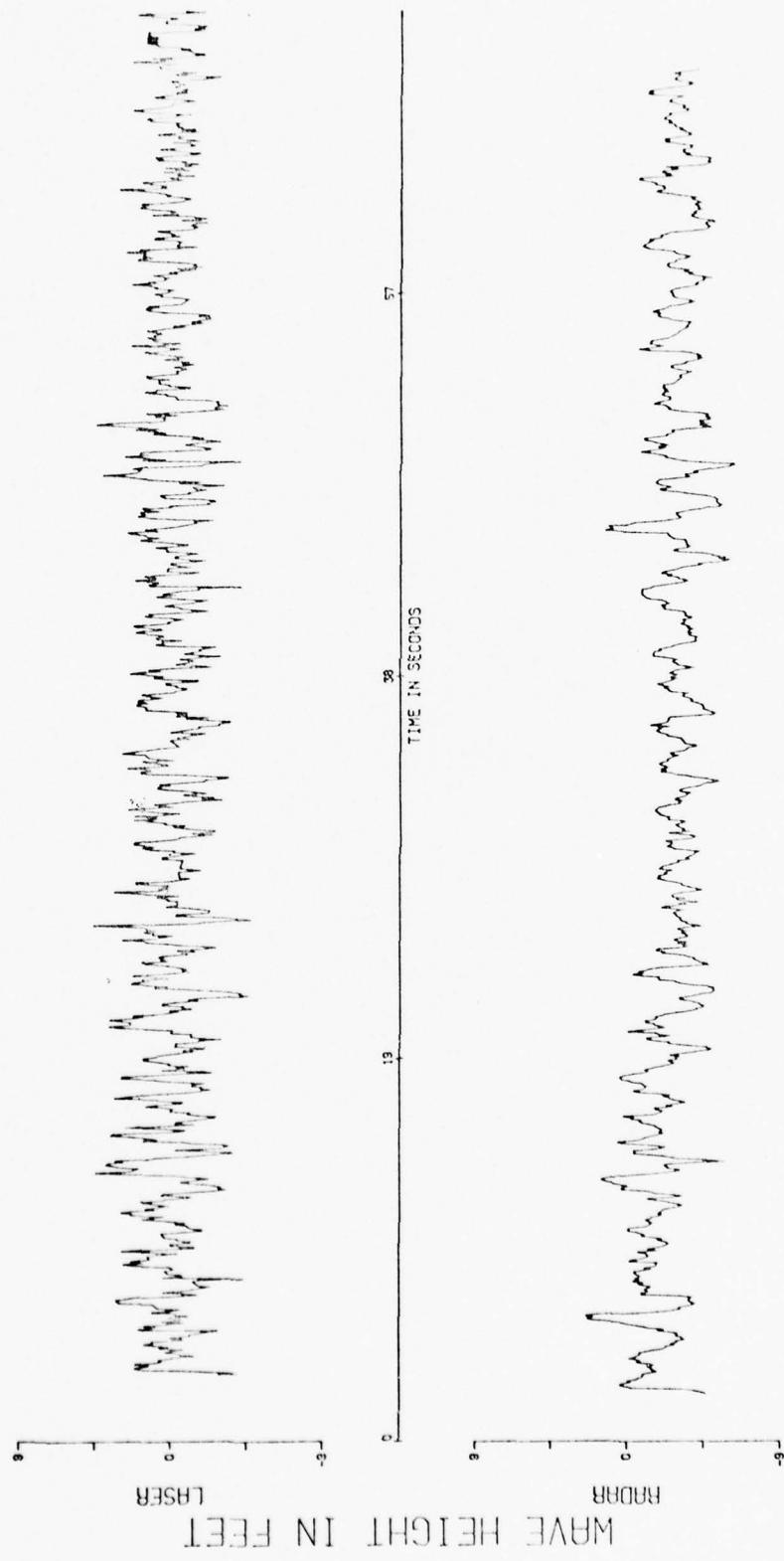


Figure 3-10

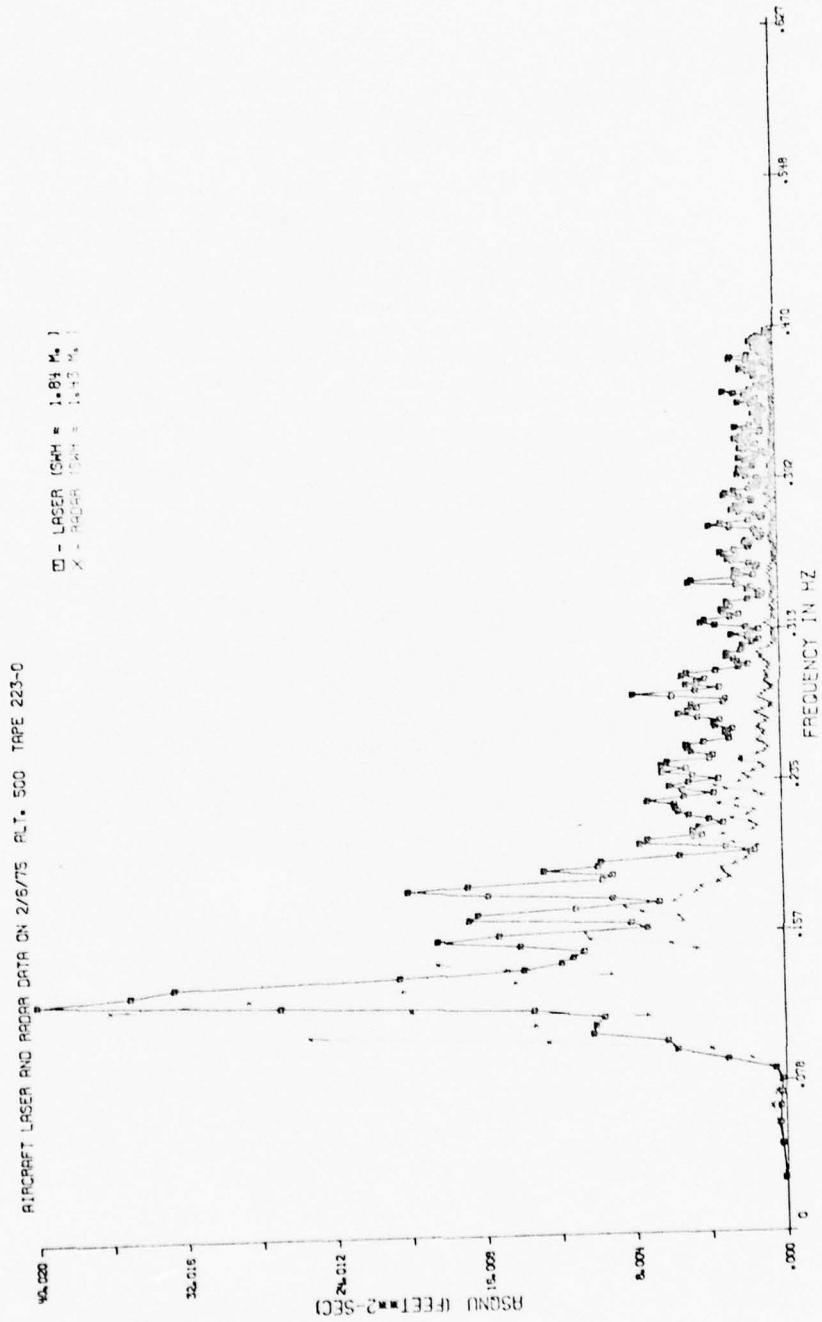
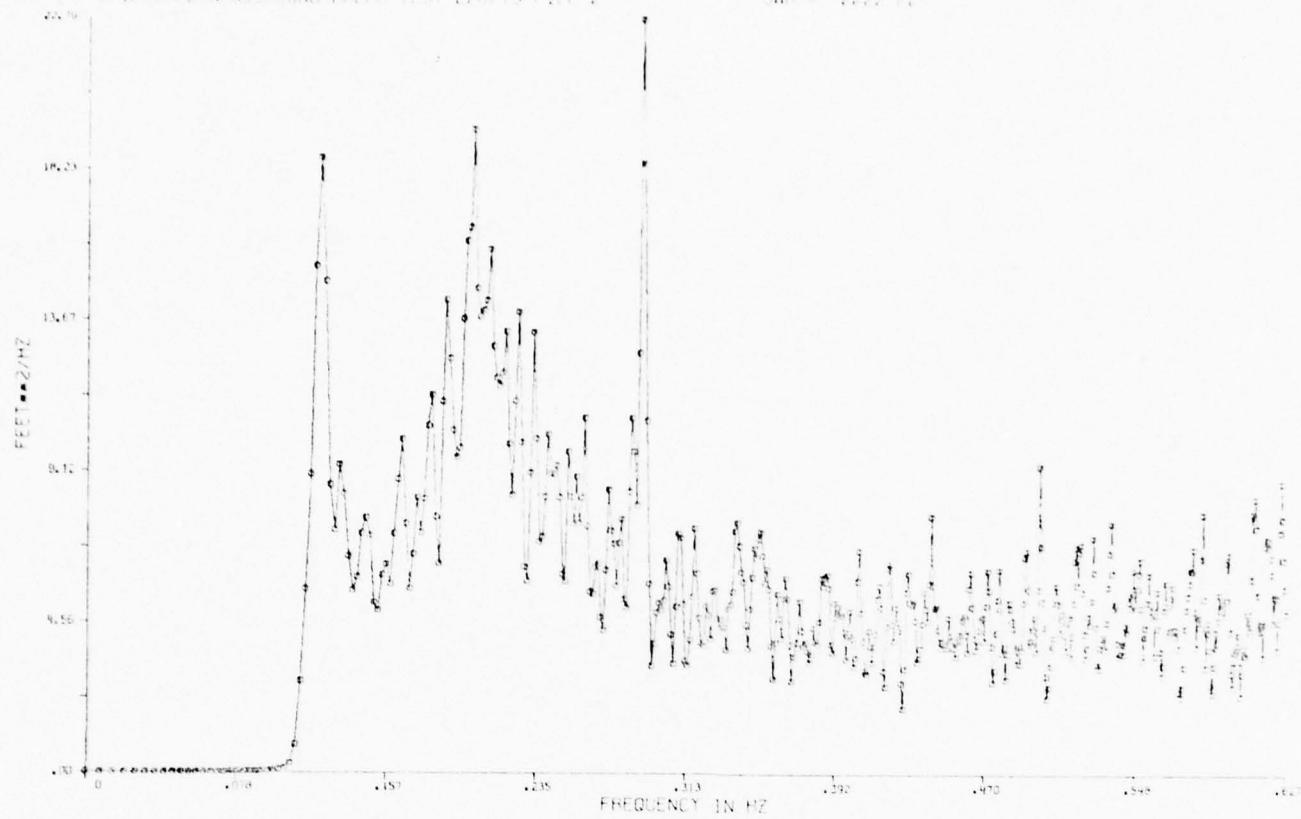


Figure 3-11

SMALL AND LARGE PENDULUM TEST 2/6/75 FILE 2

500 ft = 2.73 m



AIRCRAFT LASER AND BRORR DATA ON 2/6/75 RLT. 500 TRPE 223-0

□ - LASER (SRH = 1.84 M₀)
X - BRORR (SRH = 1.43 M₀)

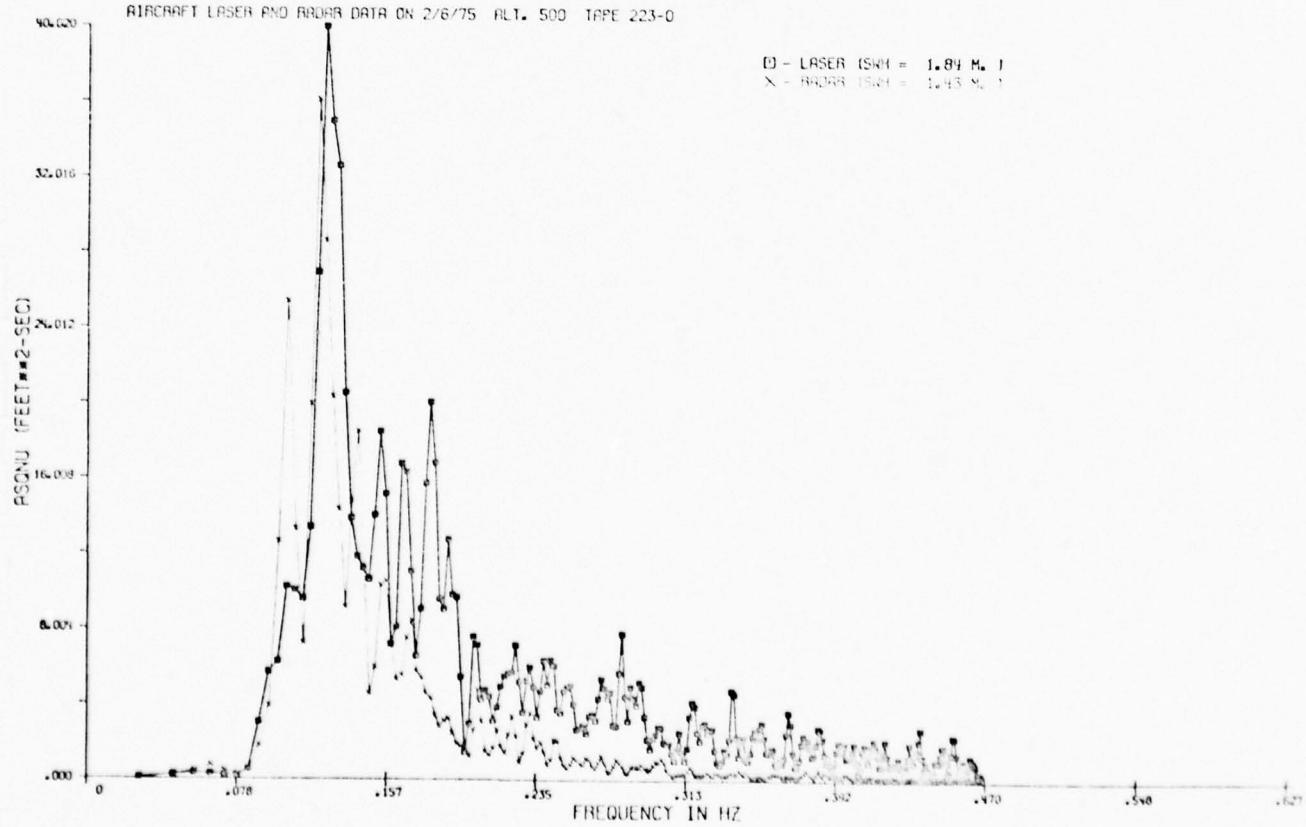


Figure 3-12

3-23

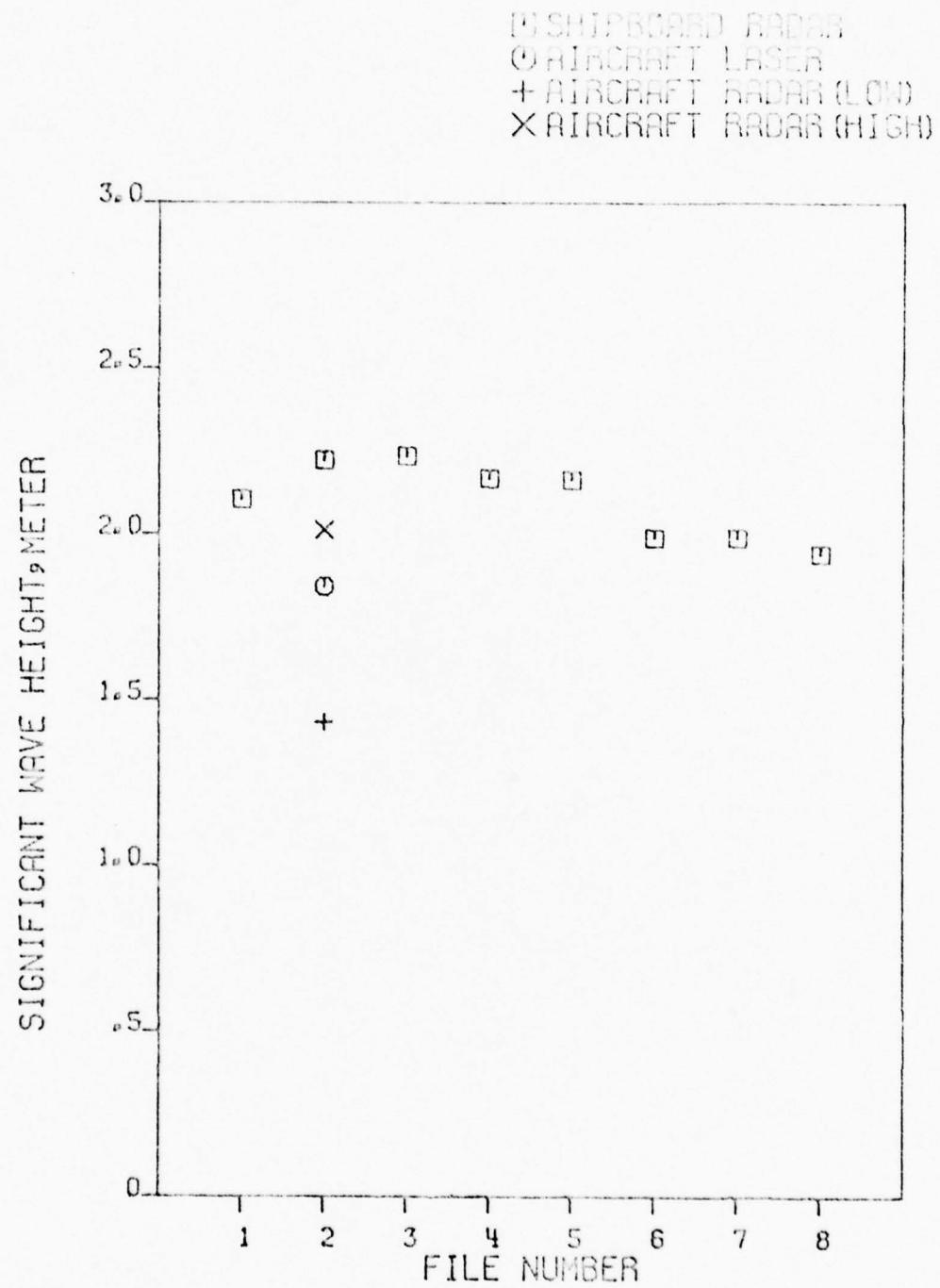


Figure 3-13

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APPENDIX A

HIGH PASS DIGITAL FILTER

Formula for the high pass filter is given by Marcel Martin in the reference by Linnette [11]. The filter used to remove the effects of $\Delta R(t) - D$ is:

$$R_f(r) = w_0 + 2 \sum_{k=1}^N w_k \cos(2\pi kr) \quad (A-1)$$

(see reference [11]) where

R_f = filter function,

r = normalized frequency which is defined by f/f_s ,

f = frequency in Hz,

f_s = sampling frequency in Hz,

w_k = the k th weighting function,

N = an integer; total number of weights is defined as

$2N + 1$.

The following are used for evaluating the terms in Equation (A-1):

$$w_k = p_k + \frac{\Delta}{2N + 1} \quad (A-2)$$

$$p_k = \left[\frac{\cos(2\pi kh)}{1 + 16h^2 k^2} \right] \left[\frac{\sin 2\pi k(r_c + h)}{\pi k} \right] \quad (A-3)$$

$$\Delta = 1 - [p_0 + 2 \sum_{k=1}^N p_k] \quad (A-4)$$

$$p_0 = 2(r_c + h) \quad (A-5)$$

$$r_c = \frac{f_c}{f_s} \quad (A-6)$$

where f_c is the ideal cutoff frequency of the filter in Hz, r_c is the normalized ideal cutoff frequency and h is the slope of the filter fall off. The filter in Equation (A-1) is designed to satisfy the following requirements:

1. the phase shift must be identically zero for all frequencies,
2. $(2N + 1)$ weights are to be used with $w_k = w_{-k}$,
3. unity gain from zero to the ideal cutoff frequency,
4. the weight calculations are optimized to achieve the conditions of a , b , and c in a least square manner with the desired filter-frequency response and considering the following:
 - a. the weighting must minimize the oscillations beyond the first zero crossing of the filter frequency-response. This is accomplished by utilizing sine function characterized by the parameter h . By selecting h sufficiently

large, the oscillations cannot exceed the preselected filter frequency fall off.

However by increasing the value of h , decreases the sharpness of the cutoff, but the sharpness of the cutoff decreases as h increases.

- b. a second iteration is made on the weight determination to insure that the filter has unit gain at zero frequency.

Since the filter must reject frequencies lower than 0.2 Hz in order to reduce the influence of $AR(t) - D$ in range variations, the following values were used to obtain that cut-off frequency:

$$N = 256$$

$$h = 0.003$$

$$r_c = 0.025.$$

The above values assume a sampling frequency of 8 Hz.

APPENDIX B

DOPPLER SHIFT CORRECTIONS

The changing pitch of the sound from, say, a fire-engine siren as it moves by at high speed is familiar to all; this effect, the Doppler effect, is one of the most obvious influences of relative motion between the source and the medium. Similarly, if a wave train of ocean waves propagating at the surface of the ocean is observed by a radar in motion, a Doppler change of frequency also will result. The Doppler effect is purely a kinematic phenomenon and can be evaluated without resorting to dynamical equations of wave motion. By kinematic argument [12], it can be shown that

$$\sigma_A = \sigma_T + \frac{\sigma_T^2}{g} v \cos \xi \quad (B-1)$$

where

σ_T = true wave frequency, $2\pi/T$ or ck_T ,

k_T = true wave number, $2\pi/\lambda_T$,

λ_T = true wavelength,

σ_A = apparent wave frequency

k_A = apparent wave number, $2\pi/\lambda_A = (\sigma_T^2/g) = (\sigma_T/v \cos \xi)$,

T = wave period,

λ_A = apparent wave length,

θ = angle between the ship's heading and wave propagation direction,

v = speed of the ship,

c = phase speed of wave component,

g = gravitational acceleration.

By employing Equation (B-1) and Jacobian transformation [13] from the apparent wave number domain to true wave frequency domain in Euclidean space yields

$$\phi(\sigma_T) = \frac{2\sigma_T}{g} \left(1 + \frac{1}{2} \frac{g}{\sigma_T v \cos\theta}\right) \psi_A(k_A) \quad (B-2)$$

where

$\phi(\sigma_T)$ = wave frequency spectrum whose argument is wave frequency σ_T , and

$\psi_A(k_A)$ = apparent wave number spectrum and

k_A is $\sigma_T^2/g + \sigma_T/v \cos\theta$.

Since data is already digitized, the equations are modified to take into account that the information is sampled at some rate and not continuous. Thus following changes are incorporated to perform the calculations.

M = total number of lags,

J = lag number,

$\Delta x = v \cos \xi$ Δt = distance between observations,

Δt = time between observations, then

$$k_A \rightarrow k_A(J) = \frac{J}{M \Delta x} = \frac{\sigma_T^2(J)}{g} + \frac{\sigma_T(J)}{v \cos \xi}$$

from Equation (A-1), and also

$$\sigma_T \rightarrow \sigma_T(J) = - \frac{g}{2v \cos \xi} + \left[\frac{g^2}{4(v \cos \xi)^2} + \frac{g \pi J}{M \Delta x} \right]^{1/2}$$

from Equation (B-2).

APPENDIX C

A	IDENT	DUMMY
	BLOCK	
	COMMON	Y(4106)•Z(4106)
	ENTRY	DUMMY
DUMMY	SLJ	88
	SLJ	DUMMY
	END	
	FINIS	

```

PROGRAM SHIPSPEC
DIMENSION Y(4106),Z(4106),ASQ2NU(513),ENUHT(513)
DIMENSION PLTARRAY(1026),IM(2),U(513),S(129)
DIMENSION FILT(257),WEIGT(257),UPHI(10),DFNC(301),IFILT(3)
DIMENSION IO(2052),RADRNG(4106),ROLL(4106),X(1026)
DIMENSION CAL(12),ICHN(12)
EQUIVALENCE (IO,Y(2053)),(RADRNG,Y),(ROLL,Z),(WEIGT,FILT)
EQUIVALENCE (X,Z(3081))
COMMON/ZAY/7
COMMON/ZBNRUFF(513)
COMMON/SF/SIZE,DELT,PLTSPACE,PL2SEC
COMMON/3/DFNC
COMMON/4/CENTROID,STANDEV,SKEWNESS,KURTOSIS
TYPE INTEGER PUNCH,PAUS,PLTON,TESTMODE
TYPE INTEGER RADRNG,ROLL,RECMNG,RADCAL,ROLCAL,FUNCTION
TYPE REAL KURTOSIS,LOOKANG,ORTNA
DATA (FILT(I),I=1,12) = 8214.,4991.,46.,10.,1432.,17251.,8214.,
1.,10.,5000.,1.)
C
C HIGHTANT = VERTICAL DISTANCE BETWEEN NANOSECOND RADAR AND STILL
C SEA LEVEL IN FEET
C NOS = NUMBER OF SAMPLES
C IM(1) = NUMBER OF POINTS IN FAST FOURIER TRANSFORM
C IM(2) = POWER OF 2
C DELT = SAMPLING TIME INTERVAL IN SECOND
C VEL = SPEED OF SHIP IN KNOT
C CUT = CUT-OFF FREQUENCY OF THE FILTER
C H = SLOPE OF WEIGHT FOR FILTER
C NUMPT = NUMBER OF POINTS IN FILTER
C LOWHI = LOW PASS FILTER = 1, HIGH PASS FILTER = 2,
C NO PASS FILTER = 3
C NF = NUMBER OF FILES TO SKIP
C NAVGL =
C LOOKANG = LOOK ANGLE FOR NANOSECOND RADAR IN DEGREE
C ORTNA = THE ANGLE BETWEEN INCOMING WIND DIRECTION AND OUTGOING
C SHIP COURSE IN DEGREE EITHER CLOCKWISE OR COUNTERCLOCKWISE
C LPHI = TITLE OF THE RUN
C N1 = NOT TO COMPUTE FILTER = 0, TO COMPUTE FILTER = 1
C N2 =
C NSKIP = 0
C LASTFILE = LAST FILE NUMBER
C PAUS = NO MACHINE PAUSE = 0, MACHINE PAUSE = 1
C PLTON = TO COMPUTE WAVE SPECTRUM = 0, NOT TO COMPUTE WAVE SPECTRUM
C = 1
C NRECSKIP = NUMBER OF FILES TO SKIP ON OUTPUT TAPE FILES
C NOWAVPLT = TO HAVE WAVE PLOT = 0, NOT TO HAVE WAVE PLOT = 1
C NPEN = CODED NUMBER FOR COLOR PENS OF THE PLOTTER
C NREC = NUMBER OF RECORDS IN THE FILE
C NWPNREC = NUMBER OF WORDS PER RECORD
C NWPDAT = NUMBER OF WORDS PER INPUT DATA IN ONE ARRAY
C NMPCAL = APPROXIMATELY LARGER THAN NUMBER OF DATA PER CALIBRATION
C LEVEL
C ICHN(1) = CHANNEL NUMBER ARRAY
C CAL = CALIBRATION
C FUNCTIONS
C 1 = WAVE SPECTRA ANALYSIS
C 2 = WRITE ON TAPE UNIT 6
C 3 = REMOVE INTEGRATION EFFECTS DUE TO ROLL
C 4 = INPUT DATA TAPE UNIT
C 5 = INPUT TAPE UNIT
C PLTSC = 0

```

```

C      PLTSPACE = SPACE INTERVAL FOR PLOT
C      MTREW = TO REWIND OUTPUT TAPE = 1, NOT TO REWIND OUTPUT TAPE = 0
C      QP =
C      TESTMODE =
C      FTRUMAX = HIGHEST FREQUENCY TO BE PLOTTED
C      DSTCCTR = DISTANCE BETWEEN RADAR AND THE CENTER OF THE SHIP
C
9001 FORMAT (3E5,F10.7,F5.0,2F10.5,4I5,F10.2)
9002 FORMAT (10A8)
9003 FORMAT (10I5)
9004 FORMAT (1H1,20X,62HTHE NEXT FILE WILL BE GOVERNED BY THE FOLLOWING
      *      RG CONTROL CARDS //,10X,19HON CONTROL CARD 1 - //,5X,3HNO5,3X,
      *      5HNC1),5X,5HNC2),6X,4HDELT,6X,1HV,8X,3HCDT,8X,1HH,4X,5HNUMPT,
      *      3X,5HLOOKHI,6X,2HNP,3X,5HNAVGL,3X,7HLOOKANG/3I8,F10.7,F8.2,2F10.5,
      *      4I8,F10.2,2X,10X,3HON CONTROL CARD 2 - LABEL IN 80 COLUMNS //,
      *      2X,10A8,2X,10X,19HON CONTROL CARD 3 - //,6X,2HNI,6X,2HN2,3X,
      *      5HNSKIP,* MAXFILE*,3X,5HPAUSE,4X,*PLOT NREP,SKP NO PLT PEND
      *      2X,10A8,2X,10X,19HON CONTROL CARD 4 - //,3X,*NUM RECK,3X,*NAPREC ORTN
      *      3A,3BX*1*7X*1*0*3X*PL2 SEC*3X*PLOT2 PT INT*6X,*MTRW*8X,5D TEST
      *      MODE FTRUMAX//2I10,F10.2,2I10,F10.0,F15.9,2I10,F10.0,2I10,F10.0,2I10
9005 FORMAT (* *0 PARITY ERROR ON ORIG INPUT TAPE*//)
9006 FORMAT (1H0)
9007 FORMAT (2I1H      MAXIMUM VALUE = *F8.2, 2I1H      MINIMUM VALUE = *
      * F8.2)
9008 FORMAT (1H1,59X,14HPOWER SPECTRUM//20X,12HTHIS RUN IS *10A8,2,15X,
      *16HAVG'D SPECTRA = *16, *50X,4DATA FACTOR = *F10.5,2,15X,16HNUMBER
      * OF LAGS = 16*50X,14HCDT = F9.4,2,15X,16HDELTA TIME = *,
      *F14.7,42X,14HSLOPE = F9.4,2,15X,16HSHP VELOCITY = F11.4,
      * 45X,14HWEIGHTS = *14,2,15X,16HVARIANCE = F11.4,45X,
      * 14HH,123 = F9.4,2,49X,4H,14H PASS FILTERED//)
9009 FORMAT (* APPARENT VARIANCE APPARENT CIRCULAR TRUE CIRC
      * TRUE FREQ TRUE FREQ * TRUE WAVE TRUE WAVE FILTER//* FR
      * 3E0X10X,4FREQ SPEC FREQ FREQ SPEC (HZ) SPECTRUM
      * NUMBER NUMBER WEIGHTS*27X,*SPECTRUM//)
9010 FORMAT (* *F8.3*2F12.4,F12.3*4F12.4,F12.3*4F12.4)
9011 FORMAT (*0 PARITY ERROR ON READ IN*)
9012 FORMAT (10X//,5X,*NUM OBS = *F11.0,2,5X,*CENTROID = *F11.4,2,5X,
      *STD DEV = *F11.4,2,5X,*SKWNESS = *F11.4,2,5X,*KURTOSIS = *F11.4)
9014 FORMAT (15,F10.2,2I15,F8.0,F10.8,2I15,F5.0,2I15,F5.0,15)
9015 FORMAT (3E8,6F12.4,2F12.6,1I0,2,1X,4F12.4,2)
9016 FORMAT (10X,19HON CONTROL CARD 5 - //,2X,*FUNCTION*,3X,
      * *1CEN*3//5X,12.5X,12.12.18) //)
9017 FORMAT (1I3I2)
C
      PAUSE 1
      CALL PLOTS(PFLTARRAY,1026+1)
      NEOP = 1MOPREV = 1FILELO = 1FILENUM = 0
      PI = 3.141592654
      DTR = PI/180.
      TWOP1 = 2.*PI
      TWOP150 = TWOP1 * 3.2
      G = 32.1729
      GNP1 = G*PI
      TWOPVG = 2.07G
      GOMTVO = G/2.
      EIGHTANT = 76.
      DEGCCR = 40.
      NDEPFC = 513
      NIDPDT = 3
      NIDPCL = 30
C
      READ (1,1) CONTROL_CARDS
C
1000 READ (1,1) CONTROL_CARDS,11,12,13,14,15,16,17,18,19,20,21,22,23,24,25,26,27,28,29,30,31,32,33,34,35,36,37,38,39,40,41,42,43,44,45,46,47,48,49,50,51,52,53,54,55,56,57,58,59,60,61,62,63,64,65,66,67,68,69,70,71,72,73,74,75,76,77,78,79,80,81,82,83,84,85,86,87,88,89,90,91,92,93,94,95,96,97,98,99,100,101,102,103,104,105,106,107,108,109,110,111,112,113,114,115,116,117,118,119,120,121,122,123,124,125,126,127,128,129,130,131,132,133,134,135,136,137,138,139,140,141,142,143,144,145,146,147,148,149,150,151,152,153,154,155,156,157,158,159,160,161,162,163,164,165,166,167,168,169,170,171,172,173,174,175,176,177,178,179,180,181,182,183,184,185,186,187,188,189,190,191,192,193,194,195,196,197,198,199,200

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200 IT = 0
READ (2,9002) LPHI
READ (2,9003) N1,N2,NSKIP,LASTFILE,PAUS,PLTON,NRECSRP,NOWAVLT,NPEN
READ (2,9014) NREC,ORTNA      ,LT,LO,PL2SEC,PLTSPACE,MTREW,OP,
* TESTMODE,TRUMAX
READ (2,9017) FUNCTION, (IHN8(I),I = 1,12)
12 = IM(1)
13 = 12/2
14 = 13+1
THETA = LOOKANG/DT
ORTNA=ORTNA/DT
R0 = HGTANT/COS(THETA)
1E6*IM(2) *NE * IM2PREV) 225,250
225 TES = -1
IM2PREV = IM(2)
C
C      NF = 0 IF NO FILES ARE TO BE SKIPPED ON INPUT TAPE
C
250 DO 300 I = 1,NF
CALL SKIPFILE(I)
IITLENUM = IITLENUM + 1
300 CONTINUE
LLENGTH = 2*NUMPT
IF (LLENGTH .GT. 512) 400,450
400 LLENGTH = 512
450 LINST = LLENGTH - 1
LLENGTH= NOS/8
LINST=LLENGTH-1
LLENGTH = 2*LLENGTH
NDATREC = NPREC/NWDAT
NLUFF = NOS/NDATREC
NPOUFF = NREC/NLUFF
NMAXCALB = NDAPCAL+20
NCAL = NOS-7
C
C      CONVERT SHIP VELOCITY FROM KNOTS TO FEET/SEC.
C
C      V = VEL*1.6878*COS(ORTNA)
C
C      SKIP NRECSRP FILES ON OUTPUT TAPE
C
DO 500 I = 1,NRECSRP
CALL SKIPFILE(I)
IITLE0 = IITLE0 + 1
500 CONTINUE
IF (V .EQ. 0.0) 502,501
501 CONST = G/(2.*V)
CONST = CONST*CONST
DX = AUSP*V*DELTA
502 NUMPT01 = NUMPT + 1
SCSD = 1
SIZE = NUMPT
NAVG = 0
AVGY = AVGNG = 0.
C
C      PRINT OUT CONTROL CARDS
C
510 PRINT 9004, 1-NOS,IM(1),IM(2),DELT,VEL,CUT,NUMPT,DELT,NE,NAVL,
*LOOKANG,LPHI,NE,NAVL,NSKIP,LASTFILE,PAUS,PLTON,NRECSRP,NOWAVLT,NPEN
*NPEN,NREC,ORTNA,LT,LO,PL2SEC,PLTSPACE,MTREW,OP,TESTMODE,
*TRUMAX
PRINT 9016, FUNCTION, (IHN8(I),I = 1,12)
NTOF = IM(1)
RENUMPT = NTOF + 0.
DO 520 I = 1,501
520 PRINT 9017, 0.

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      DO 530  I = 1114
530  ASQ2NUC1 = 0.
C
C      TEST IF FILTER IS TO BE COMPUTED.  IF(NUMP1 .GT. 0) = YES
C
C      IF(NUMP1 .EQ. 0 .OR.  NI .EQ. 0)      900,550
C
C      COMPUTE FILTER
C
C      NUMPT = NUMBER OF POINTS OVER WHICH THE AVERAGE IS TO BE TAKEN
C      CUT   = CUT OFF FREQUENCY
C      H     = SLOPE OF WEIGHTS
C
C      550  KA = NUMLPT + 1
C      CTH = CUT + H
C      FILT(KA) = 2.*CTH
C      SUMK = 0.
C      DO 700  I = 1,NUMLPT
C      P = I
C      QQ = 1. + (16.*PI*PI*P*P)
C      IF(QQ .NE. 0.)  600,575
575  FILT(I) = 0.
      GO TO 700
600  FILT(I) = (COS(2*PI*P*H)*SINE(2*PI*P*CTH))/(P1*P*QQ)
      SUMK = SUMK + FILT(I)
700  CONTINUE
      FLJO = (1. + (FILT(KA) + 2.*SUMK))/2.*NUMLPT + 1.
      DO 800  I = 1,KA
      WEIGHT(I) = FILT(I) + FLJO
800  CONTINUE
      NI = 0
900  CONTINUE
      YMAX = ZMAX = PREVMAX = -10000.
      YMIN = ZMIN = PREVMIN = 10000.
C
C      READ IN DATA
C
C      NSPECIAVG = 0
C      I1 = 1
C      I2 = 1941
C      I4 = 1242 + 1
C      I6 = I4 + 1
C      NS = I4
      FREORES = 1.*C1*DELTA*IM(C1)
      FAC0 = DELT*IM(C1)/TWOPI
      LOST = 1
      YPREV = -1000.
      TELTILENUM .EQ. 13)  910,925
910  LOST = 2501
925  ISTRT = 1
      DO 1079  TELTILENUM
      BUFR0 = IN(I1+1) - INBUFF(I1+1,NBUFF(NPRECO))
      950  TELTILENUM = 051+1050*2125*1000
1000  PRINT 9005
1050  CALL UNPKSHFT(1,CHNE,1)
1075  CONTINUE
      PRECNUM = PRECNUM + 1
      NFOR = 0
      IF(PRECNUM .GT. 10)  1400,1100
C      CALCULATING CALCULATION
1100  LDO
      YSUM = YSUM10 + ZSUM + ZSUM10 + 0.
      RADICALM = RADICALM + RADICALD + RADICALD + 0.
      SICM = 0.
      DO 1279  I = 1,LOST+1
      YSUM = YSUM + RADICALM

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ZSUM = ZSUM + ROLL(1)
YSUMSQ = YSUMSQ + RADRNG(1)*RADRNG(1)
ZSUMSQ = ZSUMSQ + ROLL(1)*ROLL(1)
IF(I .LT. 10) 1275,1150
1150 YZERO = YSUM/1
ZZERO = ZSUM/1
YSTD = SORTE((YSUMSQ - 1*YZERO*YZERO)/(I-1))
ZSTD = SORTE((ZSUMSQ - 1*ZZERO*ZZERO)/(I-1))
YTEST = 3.*YSTD
ZTEST = 3.*ZSTD
YP1 = RADRNG(I+1) - YZERO
ZP1 = ROLL(I+1) - ZZERO
YP20 = RADRNG(I+4) - YZERO
ZP20 = ROLL(I+4) - ZZERO
YP40 = RADRNG(I+7) - YZERO
ZP40 = ROLL(I+7) - ZZERO
IF(ABS(YP1) .GT. YTEST .AND. ABS(ZP1) .GT. ZTEST) 1175,1180
1175 IF(ABS(YP20) .GT. YTEST .AND. ABS(ZP20) .GT. ZTEST .AND. ABS(YP40) .GT. YTEST .AND.
* ABS(ZP40) .GT. ZTEST .AND. ABS(ZP40) .GT. ZTEST) 1300,1200
1180 IF(ABS(YP1) .GT. YTEST) 1200,1225
1200 RADRNG(I+1) = RADRNG(I)
1225 IF(ABS(YP1) .GT. ZTEST) 1250,1275
1250 ROLL(I+1) = ROLL(I)
1275 CONTINUE
RECNUMB = 0
GO TO 925
1300 IJ = I+1
JK = 0
RADCALV1 = RADCALS1 = ROLCALV1 = ROLCALS1 = 0.
DO 1315 K = IJ,NZCAL
YPCALC = RADRNG(K+1)-YZERO
ZPCALC = ROLL(K+1)-ZZERO
IF(ABS(YPCALC) .GT. YTEST .AND. ABS(ZPCALC) .GT. ZTEST)
* 1305,1320
1305 IF(K .GT. IJ) 1310,1315
1310 JK = JK+1
RADCALV1 = RADCALV1+RADRNG(K)
RADCALS1 = RADCALS1+RADRNG(K)*RADRNG(K)
ROLCALV1 = ROLCALV1+ROLL(K)
ROLCALS1 = ROLCALS1+ROLL(K)*ROLL(K)
1315 CONTINUE
1320 RADCALV2 = RADCALV1/JK
ROLCALV2 = ROLCALV1/JK
RADCALS2 = SORTE((RADCALS1+JK*RADCALV2*RADCALV2)/(JK-1))
ROLCALS2 = SORTE((ROLCALS1+JK*ROLCALV2*ROLCALV2)/(JK-1))
IF(RADCALS2 .GT. 3.*YSTD .OR. ROLCALS2 .GT. 3.*ZSTD)
* 1325,1325
1325 IF(NCHAL .GT. 0) 1326,1180
1326 J = JK
RADCALV = RADCALV*RADCALV1
RADCALSD = RADCALSD*RADCALS1
ROLCALV = ROLCALV*ROLCALV1
ROLCALSD = ROLCALSD*ROLCALS1
1326 IJ = K+1
NCOUNT = 0
NCHAL = 1
DO 1340 I = IJ,NZCAL
YP1 = RADRNG(I+1)-YZERO
ZP1 = ROLL(I+1)-ZZERO
YP20 = RADRNG(I+4)-YZERO
ZP20 = ROLL(I+4)-ZZERO
YP40 = RADRNG(I+7)-YZERO
ZP40 = ROLL(I+7)-ZZERO
NCOUNT = NCOUNT + 1
IF(YP1 .GT. YTEST .AND. ZP1 .GT. ZTEST .AND. YP20 .GT. YTEST
* ZP20 .GT. ZTEST .AND. YP40 .GT. YTEST .AND. ZP40 .GT. ZTEST)

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* 1330-1340
1330 IF (ABS(YP20) .GT. YTEST .AND. ABS(YP40) .GT. YTEST
* .AND. ABS(ZP20) .GT. ZTEST .AND. ABS(ZP40) .GT. ZTEST)
* 1340-1340
1340 CONTINUE
1345 RADCALV = RADCALV//J
ROLCALV = ROLCALV//J
RADCALSD = SQRT((RADCALSD + J)*(RADCALV*RADCALV)/(J-1))
ROLCALSD = SQRT((ROLCALSD + J)*(ROLCALV*ROLCALV)/(J-1))
1350 RACALCNT = RADCALV + YZERO
ROCALCNT = ROLCALV + ZZERO
MCHECK = 0
DO 1380 ICH = 1,12
ICHECK = ICH*ICH
IF (ICHECK .EQ. 0) 1380,1355
1355 MCHECK = MCHECK+1
IF (MCHECK .EQ. 12) 1360,1370
1360 ICH1 = ICHECK
GO TO 1380
1370 ICH2 = ICHECK
1380 CONTINUE
AMPTEST = CAL(ICH1)/1.0
DERAD = CAL(ICH1)/RACALCNT
DEROL = CAL(ICH2)/ROCALCNT
ISRT = ((I+480)/LILGTH + 1)*LILGTH
ISRT = LILGTH - MOD(IISRT+LILGTH) + IESTRT + 1
PRINT 9015, YZERO, ZZERO, RADCALV, ROLCALV, RACALCNT, DERAD,
* DEROL, ISRT, YSTD, ZSTD, RADCALSD, ROLCALSD
1400 DO 1500 I = ISRT,NOS
Y(I) = DERAD*(RADRNG(I) - YZERO)
IF (YPREV .EQ. -1000.) 1475,1425
1425 IF (ABS(Y(I) - YPREV) .GT. AMPTEST) 1450,1475
1450 Y(I) = YPREV
1475 YPREV = Y(I)
1500 Z(I) = DEROL*(ROLL(I) - ZZERO)
C DEDUCT SHIPMOTION EFFECTS DUE TO ROLL
1F (FUNCTION .EQ. 3) 1510,1530
1510 DO 1520 I = ISRT,NOS
ZRADIAN = DTR * Z(I)
VDSTRAD = RHTANT - DSTCNTR * TAN(ZRADIAN)
THETADE = ZRADIAN - THETA
RADMNNG = VDSTRAD * COS(ZRADIAN)/COS(THETADE)
1520 Y(I) = Y(I) - RADMNNG
C MESSAGE INPUT CALIBRATED DATA
1530 DO 1540 I = ISRT,NOS +LILGTH
NTIMES = 0
JEND = I + LILINST
1545 MAXPT = MINPT = AVGRNG = 0
YMAX = -10000.
YMIN = 10000.
DO 1560 I = 1900 - J, JEND
IF (Y(I) .GT. YMAX) 1550,1600
1550 YMAX = Y(I)
MAXPT = J
GO TO 1700
1600 IF (Y(I) .LT. YMIN) 1650,1700
1650 YMIN = Y(I)
MINPT = J
1700 IF (Z(I) .GT. ZMAX) 1750,1800
1750 ZMAX = Z(I)
GO TO 1900
1800 IF (Z(I) .LT. ZMIN) 1850,1900
1850 ZMIN = Z(I)
1900 AVGNSIG = AVGRNG + Y(I)
ENGAvg = AVGNSIG/LILGTH
1950 IF (ENGAvg .NE. MINPT .OR. ENGAvg .NE. MAXPT) 2050,2000

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2000 IF (MAXPT .GT. MINPT) 2020,2030
2020 Y(MAXPT) = Y(MAXPT+1)
      Y(MINPT) = Y(MINPT+1)
      GO TO 1525
2030 Y(MAXPT) = Y(MAXPT+1)
      Y(MINPT) = Y(MINPT+1)
      GO TO 1525
2050 LJUMP = 0
      NTIMES = NTIMES + 1
      IF (NTIMES .GT. 2) 2200,2075
2075 TESTYMAX = (YMAX - RNGAVG)*0.6
      TESTYMIN = (YMIN - RNGAVG)*0.6
      IF (Y(MAXPT+1) - RNGAVG .GT. TESTYMAX .OR. Y(MAXPT+1) - RNGAVG
      * .GT. TESTYMIN) 2125,2100
2100 Y(MAXPT) = (Y(MAXPT+1) + Y(MAXPT+1))/2
      LJUMP = 1
2125 IF (Y(MINPT+1) - RNGAVG .LT. TESTYMIN .OR. Y(MINPT+1) - RNGAVG
      * .LT. TESTYMIN) 2175,2150
2150 Y(MINPT) = (Y(MINPT+1) + Y(MINPT+1))/2
      LJUMP = 1
2175 IF (LJUMP .EQ. 1) 1525,2200
2200 IF (UNIT .EQ. 1) 2200,2250
2250 BUFFER OUT (ILO,1)(Y(I),Y(JEND))
      IF (YMAX .GT. PREVMAX) 2260,2270
2260 PREVMAX = YMAX
2270 IF (YMIN .LT. PREVMIN) 2280,2290
2280 PREVMIN = YMIN
2290 AVGY = AVGY + AVGRNG
      NAVG = NAVG + LENGTH
2300 CONTINUE
      IF (IILENUM .EQ. 1) STITLE .AND. RECNUMB .EQ. 312325,925
2325 NEOF = NEOF + 1
      IILENUM = IILENUM + 1
      YMAX = PREVMAX
      YMIN = PREVMIN
      IF (NEOF .GE. 2) 2300,2400
2400 PRINT 9007, YMAX,YMIN
      PRINT 9007, ZMAX,ZMIN
      PRINT 9006
      ENDFILE LO
      IFILELO = IFILELO + 1
C      OUTPUT TAPE IS READY
      IF (MTROW .EQ. 0) 2500,2450
2450 REWIND LT
      IILENUM = 0
2500 IF (FUNCTION .EQ. 2) 2510,2530
2510 PAUSE
      GO TO 100
2530 CALL BACKFILE(LO)
      CALL BACKFILE(LO)
      IF (FILELO .NE. 1) 2700,2600
2600 CALL SKIPFILE(LO)
2700 LABEL = 0
      DEINST = 1
      AVGY = AVGY/NAVG
      NRECSTRT = 1
      IMI = IMI+1
      NUMBERC = LENGTH + IMI
2800 DO 2860 I = NRECSTRT,NUMBERC,LENGTH
      IY = I
      IM = I + LENGTH
      BUFFER IN (LO,1)(Y(I),Y(IM))
2860 IF (UNIT .EQ. 1) 2810,2830
2810 PRINT 9006
2830 CONTINUE
      DOLEV = ZMAX - ZMIN + 0

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C
C      IF (PLTON .GT. 0) DO NOT COMPUTE SPECTRA
C
2926 IF (PLTON .GT. 0) 2930,2940
2930 NCOPYST = IM1 + 1
      J = 1
      DO 2935   I = NCOPYST,NUMREC
      Y(J) = Y(I)
2935 J = J + 1
      NRECSRT = J
      GO TO 2800
C
C      APPLY FILTER TO DATA (0 = NO)
C
2940 IF (NUMPT .EQ. 0) 3600,2950
2950 DO 3200   I = 1,IM1
      KI = I + NUMPT
      SUMCK = WEIGHT(KI)*Y(KI)
      DO 2960   J = 1,NUMPT
      IA = NUMPT + I - J
      IB = IA + NUMPT + 1
      IC = J - 1 + I
      SUMCK = SUMCK + WEIGHT(IA)*(Y(IC) + Y(IB))
2960 CONTINUE
C
C      LOWHI =
C      1 = LOW PASS FILTER
C      2 = HIGH PASS FILTER
C      3 = NOPASS FILTER
C
      IF (LOWHI .EQ. 1) 3100,3000
3000 SUMCK = Y(KI) - SUMCK
3100 Z(I) = SUMCK
      DCLEV = DCLEV + Z(I)
3200 CONTINUE
      DCLEV = DCLEV/IM1
C      DEDUCT STEADY STATE TERM AND CALCULATE WAVE HEIGHT DISTRIBUTION
C      FUNCTION
3600 DO 3670   M = 1,IM1
3610 X(M) = Z(M) - DCLEV
      IF (X(M) .GT. ZMAX) 3612,3614
3612 ZMAX = X(M)
      GO TO 3620
3614 IF (X(M) .LT. ZMIN) 3616,3620
3616 ZMIN = X(M)
3620 CONTINUE
      INDEX = X(M)*4.+151.5
      IF (INDEX .LT. 1) 3640,3650
3640 INDEX = 1
      GO TO 3670
3650 IF (INDEX .GT. 301) 3660,3670
3660 INDEX = 301
3670 DENC(INDEX) = DENC(INDEX) + 1
      PRINT 9007, ZMAX,ZMIN
      IF (NCWAVPLT .GT. 0) 3680,3700
3700 CALL PLOT71(FIRST,(IM1,YMIN,YMAX,ZMIN,ZMAX,LABEL))
      LAYER = 1
3800 CONTINUE
      FNTOT = FNTOT + IM1
      CALL FOURIER (X,8,IM1,IES)
      IES = 27
C      SMOOTHING BY HAMMING FUNCTION
3850 DO 3900   I = 1,IM1
3900 X(I) = X(I)*(1.0007 - X(I)*0.0007)
      IES = 0.5463333 + 0.4536667
      DO 4000   I = 2,IM1

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4000 U(J) = 0.23*(X(J-1) + X(J+1)) + 0.54*X(J)
U(14) = 0.54*X(14) + 0.46*X(16)
DO 4050 I = 1,14
4050 ASQ2NU(I) = ASQ2NU(I) + U(I)
NSPECAVG = NSPECAVG + 1
GO TO 2930
4130 IF (NOWAVPLT .GT. 0) 4180,4160
4160 IFIRST = NTOT
CALL PLOT2(IFIRST,NTOT,YMIN,YMAX,ZMIN,ZMAX,LBL1)
IF (PLTON .GT. 0) 6500,4180
4180 SPECTAVG = 2*NSPECAVG
EMAX = 0.
DO 4200 I = 1,14
U(I) = ASQ2NU(I)/SPECTAVG
EMAX = EMAX + U(I)
4200 CONTINUE
HVAR = 4.0*SORTE(EMAX)
IF (NUMLPT .EQ. 0) 4205,4210
4205 LOWHT = 3
C
C      WRITE HEADINGS
C
4210 PRINT 9008, LPHI*NSPECAVG,DERAD,IM(1),CUT,DELT,HH,V,NUMLPT,EMAX,
* HVAR,IFILT(LOWHT)
PRINT 9009
C
C      OUTPUT LOOP
C
C      REMOVE DOPPLER EFFECT
DO 4300 I = 1,14
R = 2*(I-1)
FTRU = R*REFORES
IF (V .EQ. 0,1) 4215,4213
4213 EMUHT = SORTE( CONSO + GXPI*R/(IM(1)*DX) ) + CONST
IF (I .EQ. 1) 4214,4220
4214 EMUHT = 0.
GO TO 4220
4215 EMUHT = FTRU + TWOPISO
4220 FNUHT(I) = EMUHT/TWOPI
FKT = (EMUHT*217G
ASQ1 = U(I) *FAQU
IF (V .EQ. 0,1) 4222,4221
4221 ASQMU1 = ASRF(TWODVGS*(EMUHT + CONST)*ASQ1*V)
GO TO 4223
4222 ASQMU1 = ASQ1
4223 ASQK1 = 0.
IF (EMUHT .NE. 0,1) 4225,4230
4224 ASQK1 = ASQMU1*GOD_TWZ/EMUHT
4230 ASQ2NU(I) = TWOPI*ASQMU1
IF (FTRU .GT. FTRUMAX) 4300,4240
4240 IF (I .GT. KAY) 4275,4250
4250 PRINT 9010, FTRU,U(I)*ASQ1*EMUHT*ASQMU1*FNUHT(I)*ASQ2NU(I)*FKT*
* ASQK1*REFIGE(I)
GO TO 4300
4275 PRINT 9010, FTRU,U(I)*ASQ1*EMUHT*ASQMU1*FNUHT(I)*ASQ2NU(I)*FKT*
* ASQK1
4300 CONTINUE
C      CALCULATE WAVE HEIGHT DISTRIBUTION FUNCTION FOR PLOT ONLY
4310 DO 4325 I = 5,101
IF (IDEN(EMI) .GT. 0,1) 4350,4325
4325 CONTINUE
4350 INT = NUMODE(N,4) + 1
DO 4375 MM = 4,101
M = INT - MM
IF (IDEN(EMI) .GT. 0) 4400,4375
4375 CONTINUE

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4400 IND = M + 1 - XMODE(M+4)
CENTROID = STANDEV = SKEWNESS = KURTOSIS = FNUMB = 0.
DEV = (1ST - 151)/4.
DO 5000 M = 1ST,IND
CENTROID = CENTROID + DFNC(M) *DEV
STANDEV = STANDEV + DFNC(M) *DEV**2
SKEWNESS = SKEWNESS + DFNC(M) *DEV**3
KURTOSIS = KURTOSIS + DFNC(M) *DEV**4
FNUMB = FNUMB + DFNC(M)
5000 DEV = DEV + 0.25
CENTROID = CENTROID/FNUMB
STANDEV = STANDEV/FNUMB
SKEWNESS = SKEWNESS/FNUMB
KURTOSIS = KURTOSIS/FNUMB
KURTOSIS = KURTOSIS - 4.*CENTROID*SKEWNESS + 6.*CENTROID*CENTROID*
* STANDEV - 3.*CENTROID**2
SKEWNESS = SKEWNESS - 3.*CENTROID*STANDEV + 2.*CENTROID**3
VARIANCE = STANDEV = CENTROID*CENTROID
STANDEV = SQRT(VARIANCE)
SKEWNESS = SKEWNESS/(STANDEV*VARIANCE)
KURTOSIS = KURTOSIS/(VARIANCE*VARIANCE)
PRINT 9012, FNUMB,CENTROID,STANDEV,SKEWNESS,KURTOSIS
9012 FNUMB,CENTROID,STANDEV,SKEWNESS,KURTOSIS
MIN = 1ST
MAX=IND
T BELOW = (154 - MIN)/4
T ABOVE = (MAX - 148)/4
T BELOW = T BELOW + 4*MOD(T BELOW,2)
T ABOVE = T ABOVE + 4*MOD(T ABOVE,2)
THBAR = HBAR + 1.
IF (T BELOW .GT. THBAR) 5100,5200
5100 T BELOW = THBAR
5200 IF (T ABOVE .GT. THBAR) 5300,5400
5300 T ABOVE = THBAR
5400 MIN = 151 - 4*T BELOW
MAX = 151 + 4*T ABOVE
6150 ESPECMAX = FDMAX = 0.
DO 6200 I = 1,14
IF (ASO2NU(I) .GT. ESPECMAX) 6175,6200
6175 ESPECMAX = ASO2NU(I)
6200 CONTINUE
DO 6300 I = MIN,MAX
IF (DFNC(I) .GT. FDMAX) 6250,6300
6250 FDMAX = DFNC(I)
6300 CONTINUE
CALL PLTSPEC (ASO2NU,PNUHT,LPHI,NS,I*ESPECMAX,HBAR)
CALL PLOTDF (FDMAX,T BELOW,T ABOVE,MIN,MAX)
IF (VEL .LT. 1.0) 6400,6500
6400 NF = NF/4 + 1
CALL PLTSPEC (ASO2NU,PNUHT,LPHI,NE,I*ESPECMAX,HBAR)
6500 IF (PNUHT .NE. 0 .OR. IFILENUM .EQ. 0, LASTFILE) 6600,6550
ASO2NU(LPHI(7)) = LPHI(7) + 1
IF (MOD(IFILENUM,10) .EQ. 0) 6600,100
6600 LPHI(7) = LPHI(7) - 10
LPHI(7) = LPHI(7) + 64
IF (IFILENUM .EQ. 9) 6700,100
6700 LPHI(7) = LPHI(7) - 1024
GO TO 100
6900 PAUSE
GO TO 100
9092 CALL STOPPLOT
END

```

```

SUBROUTINE PLOT2 (MIN,NOS,YMIN,YMAX,ZMIN,ZMAX,LABEL)
COMMON/SE/SIZE,PLTSPACE,DELT,TIMEMARK
DIMENSION Y(4106),Z(4106),X(1026)
EQUIVALENCE (X,Z(3081))
COMMON/AY/2
1 IF(MIN .EQ. NOS) 60,1
1 IF(LABEL .GE. 1) 4,2
2 XC = 0.
CALL SYMBOL (-1.0,0.04,0.28,17HWAVE HEIGHT (FT.),90,0,17)
CALL SYMBOL (-1.0,0.04,0.28,17HWAVE HEIGHT (FT.),90,0,17)
4 XMID = 8.0
1 IF(LABEL .GE. 1) 15,6
6 IMARK = TIMEMARK/DELT/PLTSPACE
XM = IMARK/2.
SCALAR1 = INTF(MAX1E(YMAX,ANSE(YMIN)) + 0.9)
SCALAR2 = INTF(MAX1E(ZMAX,ANSE(ZMIN)) + 0.9)
CALL PLOT (-0.1,8.0,3)
CALL PLOT (-0.0,6.0,2)
DO 10 I = 7,10
YY = I
CALL PLOT (-0.0,YY,1)
CALL PLOT (-0.1,YY,1)
CALL PLOT (-0.0,YY,1)
10 CONTINUE
CALL NUMBER (-0.5, 9.25, 0.105, SCALAR1,0.0,4HE4,0)
SCALAR1 = -SCALAR1
CALL SYMBOL (-0.5, 7.25, 0.105, 4H 0,0,0,0,4)
CALL NUMBER (-0.5, 5.25, 0.105, SCALAR1,0.0,4HE4,0)
SCALAR1 = -SCALAR1/2.
15 XX = XC
YAMP = Y(MIN)/SCALAR1+ XMID
CALL PLOT (XX,YAMP,3)
YAMP = Y(MIN+1)/SCALAR1+ XMID
XX = XX + DELT
CALL PLOT (XX,YAMP,2)
MN = MIN + 2
DO 20 I = MN,NOS
XX = XX + DELT
YAMP = Y(I)/SCALAR1 + XMID
CALL PLOT (XX,YAMP,1)
20 CONTINUE
1 IF(ZMIN .EQ. ZMAX) 55,20
30 XMID = 2.0
1 IF(LABEL .GE. 1) 45,35
35 CALL PLOT (-0.1,0.0,3)
CALL PLOT (-0.0,0.0,2)
DO 40 I = 1,4
YY = I
CALL PLOT (-0.0,YY,1)
CALL PLOT (-0.1,YY,1)
CALL PLOT (-0.0,YY,1)
40 CONTINUE
SCALAR2 = 2.*SCALAR2
CALL NUMBER (-0.5, 3.25, 0.105, SCALAR2,0.0,4HE4,0)
SCALAR2 = -SCALAR2
CALL SYMBOL (-0.5, 1.25, 0.105, 4H 0,0,0,0,4)
CALL NUMBER (-0.5, -0.25, 0.105, SCALAR2,0.0,4HE4,0)
SCALAR2 = -SCALAR2
45 XX = XC + SIZE*DELT
YAMP = X(MIN)/SCALAR1+ XMID
CALL PLOT (XX,YAMP,3)
XX = XX + DELT

```

```
YAMP = XCMINFL/SCALAR2* XMID
CALL PLOT (XY,YAMP,2)
DO 50  I = MN,NOS
XY = XY + DELT
YAMP = XCMY/SCALAR2* XMID
CALL PLOT (XY,YAMP,1)
50 CONTINUE
55 XC = XX
CALL SPACE00
RETURN
60 XX = XC
LA=XX
YMID = 1A/2
CALL PLOT (XX, 5.0+3)
XA = 1A1MARK*1MARK
DO 70  I = 1,1A+1MARK
CALL PLOT (XA, 5.0+2)
CALL PLOT (XA, 4.95+1)
CALL PLOT (XA, 5.05+1)
CALL PLOT (XA, 5.0+1)
70 XA = XA + 1MARK
CALL PLOT (0.0,0.0,-3)
RETURN
END
```

```

SUBROUTINE PLOTSPEC (YP,XP,LPHI,NOP,NP,LYMAX,HBARD)
COMMON/SE/SEIZE,DELT,PLOTSPEC,TIMENARK
DIMENSION YP(2049),XP(2049),LPHI(10)

C
C LPHI CONTAINS PLOTTING LABEL TO BE PLOTTED ON TOP OF PLOT
C MAXIMUM OF 72 CHARACTERS PLOTTED OUT
C
C YP(I) = ARRAY OF THE Y VALUES TO BE PLOTTED
C XP(I) = ARRAY OF THE X VALUES TO BE PLOTTED
C NOP = TOTAL NUMBER OF POINTS
C NP = STARTING POINT IN ARRAY FROM WHICH TO START PLOTTING
C
C
C EMAX = IMAX = 16
C XMIN = EMAX/XP(NOP)
C
30 SCALE = (INTE(100,LYMAX+0.99))/100.
CALL PLOT (0.0, -0.1, 3)
CALL PLOT (0.0, 1.0, 2)
DO 10 I = 1,9
Y = I
CALL PLOT (-0.1, Y, 1)
CALL PLOT (0.0, Y, 1)
CALL PLOT (0.0, Y+1, 1)
10 CONTINUE
SWH = 0.3048HHAQ
CALL SYMBOL (0.1,10.2,0.175,LPHI+0,0,72)
CALL SYMBOL (9.0,10.2,0.175,14HS0M = -M+0,0,14)
CALL NUMBER (9.9,10.2,0.175,SWH,0.0,4HES,2)
YY = 9.95
FLABEL = SCALE
DFCR = SCALE/5.
DO 20 I = 1,6
CALL NUMBER (-0.7,YY,0.105,FLABEL,0.0,4HE6,2)
FLABEL = FLABEL - DFCR
YY = YY - 2.0
10 CT, 40, 41, 15, 20
40 CALL SYMBOL (-0.9,-0.275,0.175,10HEET*2.0H,9.0,0,10)
20 CONTINUE
EMID = EMAX/2.
L = -1
DO 30 I = NP,NOP
YVAL = YP(I)/SCALE*10.
XVAL = XP(I)/XMIN
CALL SYMBOL (XVAL,YVAL, -0.07,0.0,0,1)
30 L = L + 1
X = EMAX
CALL PLOT (X,-0.1,3)
CALL PLOT (X, 0.0,2)
DO 40 I = 1,IMAX-1
X = IMAX - I - 1
CALL PLOT (X, 0.0,1)
CALL PLOT (X,-0.1,1)
CALL PLOT (X, 0.0,1)
40 CONTINUE
CALL PLOT (-0.1,0.0,1)
X = -0.1
IMAX = EMAX + 1
DO 50 I = 1,IMAX-1
XX = I - 1
CPTL = XN*YMIN
CALL NUMBER (X,XX,CPTL,0.0,4HES,2)
50 CT, 41, 42, 15, 20
41 CALL SYMBOL (IMAX-1.0,0.0,0.175,10HEET*2.0H,9.0,0,10)

```

```
DO X = X + 2.0
CALL PLOT (FMAX+3.0,0.0,-3)
RETURN
END
```

```

SUBROUTINE PLOTDF(DM, T BELOW, T ABOVE, MIN, MAX)
COMMON/3/DF
COMMON/4/CENTROID, STANDEV, SKWNESS, KURTOSIS
TYPE REAL KURTOSIS
DIMENSION DF(301)
DFMAX = (XF*XF(DM + 99.0/100)*100
NPOINTS = (T ABOVE + T BELOW)*4
INCHES = (NPOINTS + 4)/8
HALF = INCHES/2. - 0.275
CALL PLOT (0.0,-0.1,3)
CALL PLOT (0.0, 1.0,2)
YY = 1.0
DO 10  T = 1.0
CALL PLOT (-0.1,YY,1)
CALL PLOT (0.0,YY,1)
YY = YY + 1.0
10 CALL PLOT (0.0,YY,1)
CALL SYMBOL (0.2,10,20,0.175,25H01ST, FN, OF WAVE HEIGHTS,0.0,25)
CALL SYMBOL (0.2,9.75,0.175,11HCENTROID = 0.0,11)
CALL NUMBER (1.85,9.75,0.175,CENTROID,0.0,4HFB,4)
CALL SYMBOL (0.2,9.50,0.175,11HSTAN DEV = 0.0,11)
CALL NUMBER (1.85,9.50,0.175,STANDEV,0.0,4HFB,4)
CALL SYMBOL (0.2,9.25,0.175,11HSKEWNESS = 0.0,11)
CALL NUMBER (1.85,9.25,0.175,SKWNESS,0.0,4HFB,4)
CALL SYMBOL (0.2,9.00,0.175,11HKURTOSIS = 0.0,11)
CALL NUMBER (1.85,9.00,0.175,KURTOSIS,0.0,4HFB,4)
XX = -0.55
YY = 0.93
TDF = DFMAX
TDFDEC = TDF/5
DO 20  T = 1.0
CALL NUMBER (XX,YY,0.14,TDF,0.0,7H14)
IF (T .EQ. 3) 14,16
14  CALL SYMBOL (-0.7,4.575,0.175,6HNUMBER,0.0,6)
15  YY = YY + 2.0
TDF = TDF + TDFDEC
20  CONTINUE
CALL PLOT (-0.1,0.0,3)
CALL PLOT (0.0,0.0,2)
XX = 1.0
DO 30  T = 1.0,INCHES
CALL PLOT (XX, 0.0,1)
CALL PLOT (XX,-0.1,1)
CALL PLOT (XX, 0.0,1)
30  XX = XX + 1.0
IX = T ABOVE
YY = INCHES - 0.15
INCHES = INCHES + 1
DO 40  T = 1,INCHES,2
CALL NUMBER (XX,-0.25,0.14,IX,0.0,2H13)
IF (XX .EQ. 0 .OR. HALF .EQ. 0 .OR. XX .EQ. 0 .OR. HALF) 34,36
34  CALL SYMBOL (HALF,0.5,0.175,4HFFET,0.0,4)
36  IX = IX - 4
40  XX = XX + 2.0
XX = 0.0
L = -1
DO 50  T = MIN,MAX
YY = (DF(T)) / DFMAX * 10
CALL SYMBOL (XX,YY,0.0,0.0,0)
L = L + 1
50  XX = XX + 0.125
60  CONTINUE

```

```
XINCHES = INCHES + 4
CALL PLOT (XINCHES,0,0,-3)
RETURN
END
```

```

C SUBROUTINE FOURIER (A,S,M,IES)
C
C THIS ROUTINE PERFORMS AN ANALYSIS OF 2**M POINTS BY FIRST DOING
C AN ANALYSIS OF 2**M/2 COMPLEX POINTS AND THEN ARRANGING THE RESULTS
C
C ARGUMENTS
C 1. A = REAL DATA ARRAY = OF DIMENSION 2**M + 2
C 2. S = SIN/COS TABLE = DIMENSION 2**M(M-3)
C 3. M = EXPONENT OF 2 = SIZE OF REAL ARRAY
C 4. IES = -1 FOR FIRST TIME, -2 THEREAFTER
C
C DIMENSION A(11,50)
C N = 2**M-1
C CALL HARMONIA,S,M-1,IES,IERRO
C MERGE 2 N-POINT ANALYSIS INTO 1 2N-POINT ANALYSIS
C NHALF = N/2
C NTWO = N/2 + 4
C X = X0 = COS(3.1415926536/FLOAT(N))
C Y = Y0 = SIN(3.1415926536/FLOAT(N))
C 00 1000 K2 = 4,N/2
C K1 = K2 - 1
C N2 = NTWO - K2
C N1 = N2 - 1
C BK1 = A(K1) + A(N1)
C BK2 = A(K2) + A(N2)
C BN1 = A(K2) + A(N2)
C BN2 = A(K1) + A(N1)
C XBN1 = X*BN1
C XBN2 = X*BN2
C YBN1 = Y*BN1
C YBN2 = Y*BN2
C
C A(K1) = .5 *(BK1 + XBN1 + YBN1)
C A(K2) = .5 *(BK2 + XBN2 + YBN2)
C A(N1) = .5 *(BK1 - XBN1 - YBN1)
C A(N2) = .5 *(BK2 - XBN2 - YBN2)
C Q = X*X0 - Y*Y0
C Y = Y*Y0 + X*Y0
C 1000 X = Q
C
C COMPLEX ELEMENT A(N)
C A(2*N+1) = (A(1) - A(2))*i
C A(2*N+2) = 0.0
C
C COMPLEX ELEMENT A(0)
C A(1) = .5*(A(1)+i*A(2))
C A(2) = 0.0
C
C COMPLEX ELEMENT A(N/2)
C A(NFLY) = A(NFLY)
C A(NFLY) = A(NFLY)
C
C RETURN
C END

```

SUBROUTINE HARMON(A,S,M,IFS,IFERR)
 DIMENSION A(1),S(1)
 C HARM, ONE-DIMENSIONAL BASIC FORTRAN VERSION, J.W.COOLEY HARM 001
 C MODIFIED TO RUN ON CDC 3800 AND TO ANALYZE UP TO 2**14 NUMBERS. HARM 002
 C
 C
 C DOES EITHER FOURIER SYNTHESIS, I.E., COMPUTES COMPLEX FOURIER SERIESHARM 009
 C GIVEN A VECTOR OF N COMPLEX FOURIER AMPLITUDES, OR, GIVEN A VECTOR HARM 010
 C OF COMPLEX DATA X DOES FOURIER ANALYSIS, COMPUTING AMPLITUDES. HARM 011
 C A IS A COMPLEX VECTOR OF LENGTH N=2**M COMPLEX NOS. OR 2**N REAL HARM 012
 C NUMBERS. A IS TO BE SET BY USER. HARM 013
 C M IS AN INTEGER SUCH THAT 0<=M<=14 SET BY USER. HARM 014
 C S IS A VECTOR S(J)=SIN(2*PI*J/NP) J=1,2,...,NP/4-1. HARM 015
 C COMPUTED BY PROGRAM. HARM 016
 C IFS IS A PARAMETER TO BE SET BY USER AS FOLLOWS- HARM 017
 C IFS=0 TO SET NP=2**M AND SET UP SINE TABLE S. HARM 018
 C IFS=1 TO SET N=NP=2**M, SET UP SIN TABLE, AND DO FOURIER HARM 019
 C SYNTHESIS, REPLACING THE VECTOR A BY HARM 020
 C
 C X(J)= SUM OVER K=0,N-1 OF A(K)*EXP(2*PI*I/N)**(J,K), HARM 021
 C J=0,N-1, WHERE I=SORT(-1). HARM 022
 C THE X'S ARE STORED WITH RE X(J) IN CELL 2*NJ+1 HARM 023
 C AND IM X(J) IN CELL 2*NJ+2 FOR J=0,1,2,...,N-1. HARM 024
 C THE A'S ARE STORED IN THE SAME MANNER. HARM 025
 C
 C IFS=+1 TO SET N=NP=2**M,SET UP SIN TABLE, AND DO FOURIER HARM 026
 C ANALYSIS, TAKING THE INPUT VECTOR A AS X AND HARM 027
 C REPLACING IT BY THE A, SATISFYING THE ABOVE FOURIER SERIES. HARM 028
 C IFS=+2 TO DO FOURIER SYNTHESIS ONLY, WITH A PRE-COMPUTED S. HARM 029
 C IFS=-2 TO DO FOURIER ANALYSIS ONLY, WITH A PRE-COMPUTED S. HARM 030
 C IFERR IS SET BY PROGRAM TO- HARM 031
 C =0 IF NO ERROR DETECTED. HARM 032
 C =1 IF M IS OUT OF RANGE, OR, WHEN IFS=+2,-2, THE HARM 033
 C PRE-COMPUTED S TABLE IS NOT LARGE ENOUGH. HARM 034
 C =-1 WHEN IFS =+1,-1, MEANS ONE IS RECOMPUTING S TABLE HARM 035
 C UNNECESSARILY. HARM 036
 C
 C NOTE- AS STATED ABOVE, THE MAXIMUM VALUE OF M FOR THIS PROGRAM HARM 037
 C ON THE IBM 7094 IS 13, ON 360 MACHINES HAVING GREATER STORAGE HARM 038
 C CAPACITY, ONE SHOULD CHANGE THIS LIMIT BY REPLACING 13 IN HARM 039
 C STATEMENT 3 BELOW BY LOG2 N, WHERE N IS THE MAX. NO. OF HARM 040
 C COMPLEX NUMBERS ONE CAN STORE IN HIGH-SPEED CORE. HARM 041
 C
 C IF THE CAPACITY OF HARM IS TO BE INCREASED, ONE MUST HARM 042
 C ALSO ADD MORE DO STATEMENTS TO THE BINARY SORT ROUTINE HARM 043
 C FOLLOWING STATEMENT 24, AND CHANGE THE EQUIVALENCE STATEMENTS HARM 044
 C FOR THE K'S. HARM 045
 C
 C DIMENSION K(14)
 C EQUIVALENCE (K(14),K14),(K(13),K21),(K(12),K3),(K(11),K4) HARM 046
 C EQUIVALENCE (K(10),K5),(K(9),K6),(K(8),K7),(K(7),K8) HARM 047
 C EQUIVALENCE (K(6),K9),(K(5),K10),(K(4),K11),(K(3),K12) HARM 048
 C EQUIVALENCE (K(2),K13),(K(1),K14),(K(1),K2) HARM 049
 C IF(M>2,2,3)
 3 IF(M>LE-14) 5,6,2 HARM 050
 2 IFERR=1 HARM 051
 1 RETURN HARM 052
 6 IF(IFERR<0) HARM 053
 N=2**M HARM 054
 7 IF(IFERR<0) 8,9,10 HARM 055
 C
 C 8 IF(IFERR<0) 9,10,20,20,10 HARM 056
 C
 C 9 IF (A(1)=0) GO TO 10 HARM 057
 C
 C 10 IF (N>NP) 11,0,20,12 HARM 058

```

12 IFERR=1
      GO TO 200
C      SCRAMBLE A, BY SANDE'S METHOD
20 K(1)=2*N
      DO 22 L=2,M
22 K(L)=K(L-1)/2
      DO 24 L=M,13
24 K(L+1)=2
C      NOTE EQUIVALENCE OF K(L) AND K(14-L)
C      BINARY SORT-
      IJ=2
      J1=2
25 DO 30 J2=J1+1,K1
      DO 30 J3=J2+1,K2
      DO 30 J4=J3+1,K3
      DO 30 J5=J4+1,K4
      DO 30 J6=J5+1,K5
      DO 30 J7=J6+1,K6
      DO 30 J8=J7+1,K7
      DO 30 J9=J8+1,K8
      DO 30 J10=J9+1,K9
      DO 30 J11=J10+1,K10
      DO 30 J12=J11+1,K11
      DO 30 J13=J12+1,K12
      DO 30 J14=J13+1,K13
      IF(IJ,J11,29,30,30)
28 T=A(IJ+1)
      A(IJ+1)=A(J1+1)
      A(J1+1)=T
      T=A(IJ)
      A(IJ)=A(J1)
      A(J1)=T
30 IJ=IJ+2
      J1=J1+2
      IF(K1-IJ)31+25,25
31 IF(IJSY)32,2,36
C      DOING FOURIER ANALYSIS, SO DIV. BY N AND CONJUGATE.
32 FN = FLOAT(N)
      DO 34 I=1,N
      A(2*I-1) = A(2*I-1)/FN
34 A(2*I)=A(2*I)/FN
C      SPECIAL CASE- L=1
36 DO 40 I=1,N,2
      T = A(2*I-1)
      A(2*I-1) = T + A(2*I+1)
      A(2*I+1) = T + A(2*I+1)
      T=A(2*I)
      A(2*I) = T + A(2*I+2)
40 A(2*I+2) = T + A(2*I+2)
      IF(M-1)2+1,450
C      SET FOR L=2
50 LEXP1=2
C      LEXP1=2** (L-1)
      LEXP=2
C      LEXP=2** (L-1)
      NPL = 2*PI*LN2
C      NPL = 80* 1.884+1
      DO 130 L=2,M
C      SPECIAL CASE- L=2
      DO 90 L2=2,M,LEXP
      L1=L + LEXP1
      L2=L2+LEXP1
      L=L+LN2*LN2
      T=A(IJ)
      A(IJ) = T + A(IJ+1)
      A(IJ+1) = T + A(IJ+1)
      IF(M-1)131,130

```

```

T = A(1)
A(1) = T+A(2)
A(2) = T-A(1)
T = -A(3)
T1 = A(3)-1
A(13-1) = A(11-1) - T
A(13-1) = A(11-1) - T1
A(11-1) = A(11-1) + T
80 A(11-1) = A(11-1) + T1
IF(L=2) 120,120,90
90 KLAST=N2+LEXP
JJ=NPL
DO 110 J=4,LEXP1+2
N0JJ=NT-JJ
UR=S(NPJJ)
UI=S(JJ)
ILAST=J+KLAST
DO 100 I= J+ILAST+LEXP
I1=I+LEXP1
I2=I1+LEXP1
I3=I2+LEXP1
T=A(12-1)*UR-A(12)*UI
T1=A(12-1)*UI+A(12)*UR
A(12-1)=A(11-1)-T
A(12-1)=A(11-1) - T1
A(11-1)=A(11-1)+T
A(11) =A(11)+T1
T=A(13-1)*UI-A(13)*UR
T1=A(13-1)*UR-A(13)*UI
A(13-1)=A(11-1)-T
A(13) =A(11-1)-T1
A(11-1)=A(11-1)+T
100 A(11) =A(11) +T1
C END OF I LOOP
110 JJ=NPL
C END OF J LOOP
120 LEXP1=2*LEXP1
LEXP = 2*LEXP
130 NPL=NPL/2
C END OF L LOOP
IF(L>ES)145,24
CC DOING FOURIER ANALYSIS. REPLACE A BY CONJUGATE.
145 DO 150 I=1,N
150 A(2*I) =-A(2*I)
GO TO 1
C RETURN
C MAKE TABLE OF S(J)=SIN(2*PI*I/JNP), J=1,2,...,NT-1,NT=NP/4
200 N0N
M0M
NT=N/4
MT=M+2
IF(MT) 260,260,205
205 THE=TA=78539816.34
C THE=PI/2**((L-1)) FOR L=1
J1DE = NT
C J1DE = 2**((MT-1)) FOR L=1
J1DE = NT/2
C J1DE = 2**((MT-1)) FOR L=1
S(J1DE) = SIN(THETA)
IF (MT>2) J1DE=220,220
220 DO 230 I=1,MT
THE=TA=THE*TA*
J1DE = J1DE+1
J1DE = J1DE/2
J1DE = J1DE/2
S(J1DE)=SIN(THETA)
230

```

JC1=NT-JD1F HARM 194
S(JC1)=COS(THETA) HARM 195
JLAST=NT-JSTEP2 HARM 196
IF (JLAST-JSTEP)250,230,230 HARM 197
230 DO 240 J=JSTEP+JLAST+JSTEP HARM 198
JC=NT-J HARM 199
JD=J+JD1F HARM 200
240 S(JD)=S(J)*S(JC)+C(JD1F)*S(JC) HARM 201
250 CONTINUE HARM 202
260 IF (JES)20,1,20 HARM 203
END

```
SUBROUTINE UNPKSHIP (INCHR,I)
COMMON/A/RADRNG(4106),ROLL(4106)
COMMON/B/NPUFF(513)
DIMENSION INCHR(12),NBUFF1(3),N1(2)
TYPE INTEGER RADRNG,ROLL
DATA (M1=3777B)+(M2=2000B)+(M3=7777777774000B)
I1=0
DO 9 K1=1,513,3
I1=I1+1
I2=(I1-1)*171+I1
DO 4 J=1,3
NBUFF1(J)=NPUFF(K1+J-1)
4 CONTINUE
I3=0
DO 8 K=1,12
IF (INCHR(K).NE.0) I1=8
I I3=I3+1
IWORD=INCHR(K)-117441
INDEX=128*IWORD*4+INCHR(K)+1
NTSENBUFF(IWORD)/288 INDEX
NTSENTS,AND,M1
NTSIGNENTS,AND,M2
IF (NTSIGN.EQ.0) 6,7
6 NTS=NTS+1
NTSENTS,OR,M3
7 CONTINUE
N1(I3)=NTS
8 CONTINUE
RADRNG(12)=N1(1)
ROLL(12)=N1(2)
9 CONTINUE
RETURN
END
```

METRIC CONVERSION FACTORS

Approximate Conversions to Metric Measures

میں اسی سے 250 میل کی مسافت کو 25 کی سرعت پر 100 سے 100 کی سرعت پر کوئی تاثر نہیں ملے۔

ANALYTICAL CONSIDERATIONS FROM NATURE MEASURES

Approximate Conversions from Metric Measures									
Symbol	When You Know	Multiply by	To Find	Symbol					
<u>LENGTH</u>									
mm	millimeters	0.04	inches	in					
cm	centimeters	0.4	inches	in					
m	meters	3.3	feet	ft					
m	meters	1.1	yards	yd					
km	kilometers	0.6	miles	mi					
<u>AREA</u>									
cm ²	Square centimeters	0.16	Square inches	in ²					
m ²	Square meters	1.2	Square yards	yd ²					
km ²	Square kilometers	0.4	Square miles	mi ²					
ha	hectares (10,000 m ²)	2.5	acres	acres					
<u>MASS (WEIGHT)</u>									
g	grams	0.035	ounces	oz					
kg	kilograms	2.2	pounds	lb					
t	tonnes (1,000 kg)	1.1	short tons	sh tn					
<u>VOLUME</u>									
ml	milliliters	0.03	fluid ounces	fl oz					
l	liters	2.1	pints	pt					
l	liters	1.06	quarts	qt					
m ³	cubic meters	35	gallons	gal					
m ³	cubic meters	1.3	cu ft	cu ft					
<u>TEMPERATURE (exact)</u>									
°C	Celsius temperature	9/5 (then add 32)	Fahrenheit temperature	°F					
			100°	212°					
			50°	160°					
			40°	140°					
			30°	120°					
			20°	100°					
			10°	80°					
			0°	60°					
			-10°	40°					
			-20°	20°					
			-30°	0°					
			-40°	-20°					
			-50°	-40°					

Days	Cases
0	0
10	10
20	25
30	40
40	55
50	65
60	70
70	75
80	78
90	80
100	82

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Technical Report Documentation Page

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ABSTRACT

A microwave shipboard wave-height radar sensor for measuring ocean wave spectra, developed by the Naval Research Laboratory, was installed on the containership *S.S. NOZEM*, February, 1975. The sensor's performance, design, and analysis of data for one data run are discussed. The radar system has a 3 centimeters wavelength, 2 nanoseconds pulse width, 100 watts of peak transmitted power, 10,000 pulse per second repetition rate, 2-foot parabola antenna diameter, 7 decibel receiver noise figure, 100 pulses per second equivalent pulse processing rate, and a 1-foot resolution. Results are in reasonable agreement with airborne measurements. Areas for improving the system are also discussed.

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